

A Technical Discussion of TDR
Techniques, S-parameters, RF
Sockets, and Probing Techniques
for
High Speed Serial Data
Designs

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Agenda

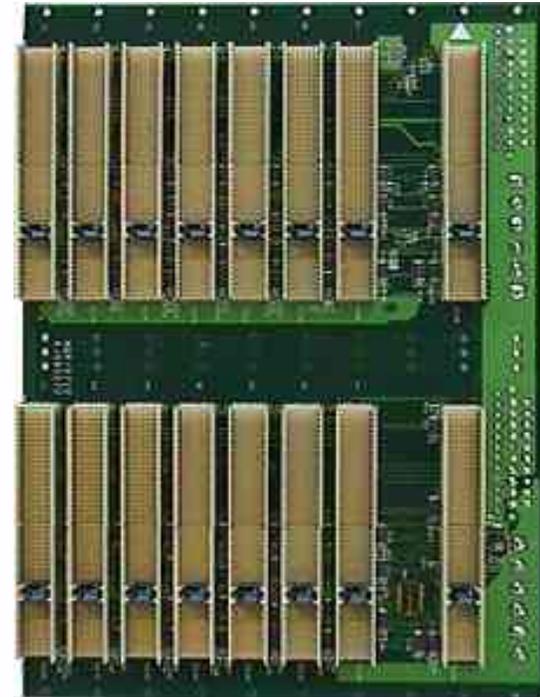
Technical Discussion of TDR Techniques, S-parameters, Sockets, and Probing Techniques

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- *Why do we need TDR?*
- *What is TDR?*
 - *TDR Block Diagram*
- *What is TDT?*
- *Differential TDR/TDT*
- *Basic TDR measurements*
- *Sources of L/C Discontinuities (sockets, connectors, etc.)*
- *Using TDR to Locate Discontinuities (open/short faults)*
- *Enhancing TDR Measurement through Calibration (Normalization)*
- *TDR Rise Time considerations*
- *What are S-parameters?*
 - *Impedance In the Time and Frequency Domains*
 - *Comparison of TDR derived S-parameters vs. VNA*
- *TDR Probes*

Technology Developments in Data Transmission...

- Faster Clock and Data rates in computers, communications and network systems
- New standards are all in the Gbps range
 - *PCIe Gen(2) 5 Gb/s*
 - *SATA Gen(3)/SAS2*
 - *10 Gigabit Ethernet*
- *Faster rates imply sharper pulses with faster rise times*
- *Impedance - controlled transmission lines used to carry Serial transmission*



High speed backplane

High speed data link requirements...

- Factors originally ignored at slower data rates, now become critical design considerations - *Crosstalk, ISI*

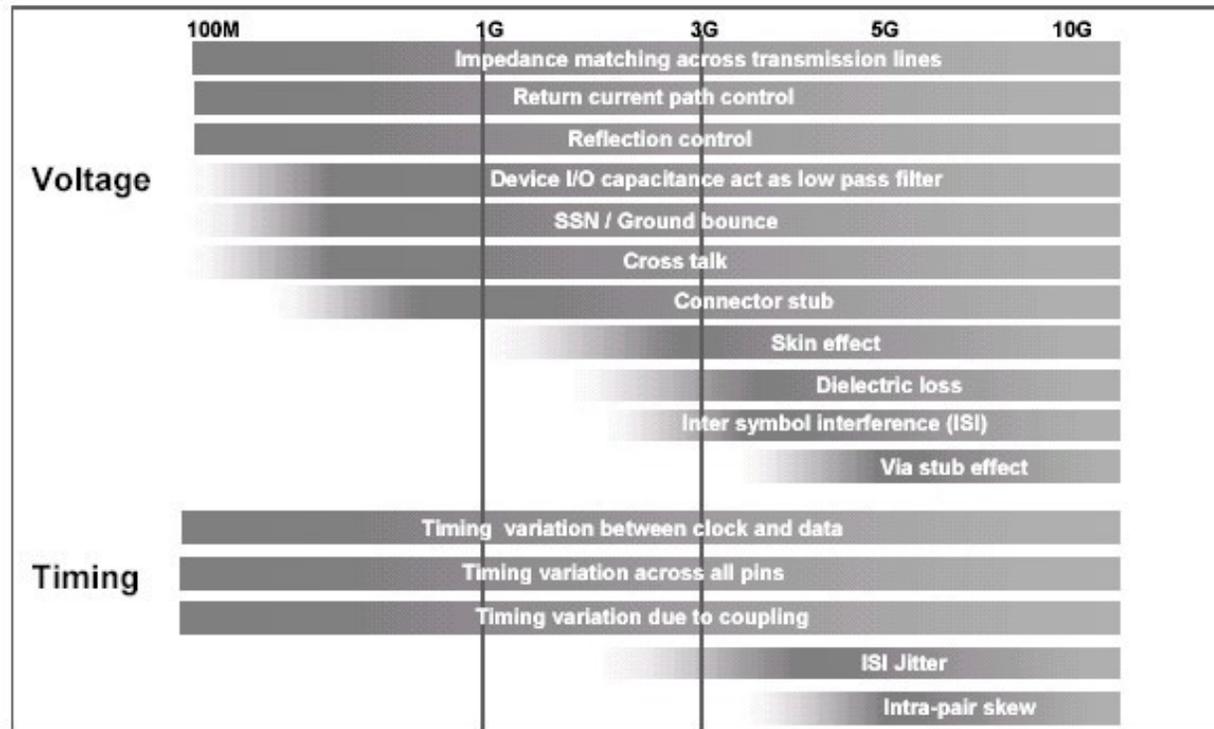


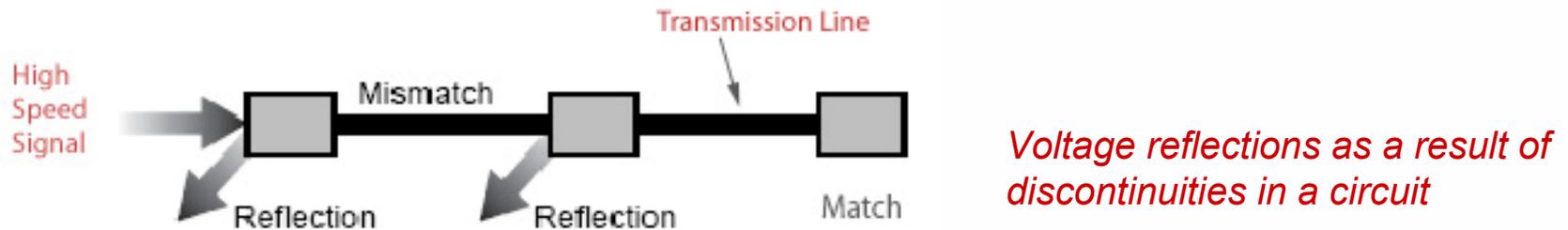
Figure shows the key timing and voltage related impairments that must be addressed as data rates increase

- Transmission line quality critical for data transfer at higher rates
- TDR* analysis now part of signal integrity measurements of high speed circuits along with *Jitter*

What is TDR & TDT

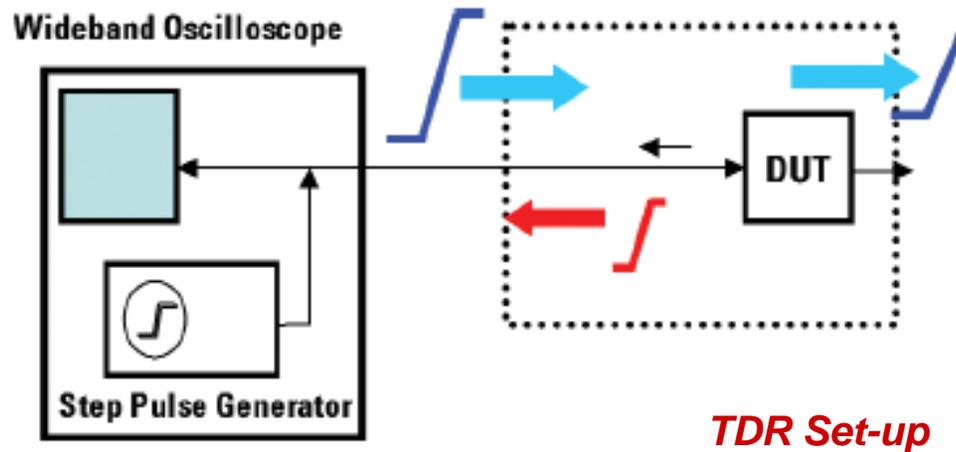
What is TDR?

- **TDR** - Time Domain Reflectometry (TDR)



- Measure of Reflections on an *applied step pulse* to a Device Under Test (DUT) - *PCB traces, cables, etc.*
- Powerful tool for measuring *Impedance* through a circuit
Also for measuring *Discontinuities* that cause reflections and the *Distance* of the discontinuities.
- Shows the effects of poor connections, mismatched traces, circuit discontinuities etc.

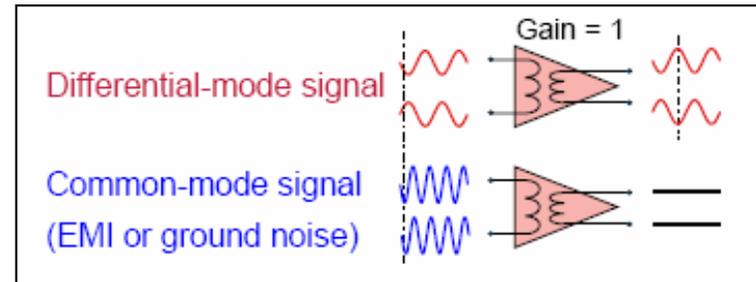
TDR Block Diagram



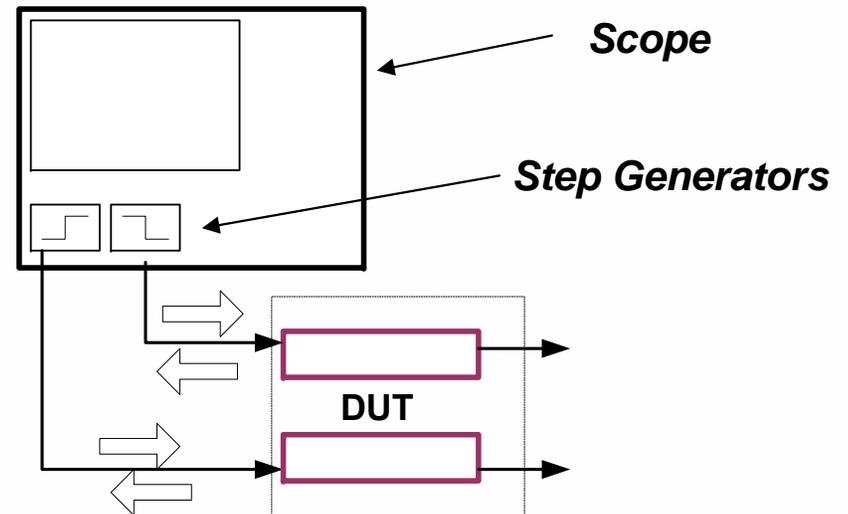
- A pulse generator is used to provide an incident step pulse
- Voltage Reflection from the Device Under Test (DUT) is measured by the scope
- Shape of the measured Reflection helps determine the type of discontinuity and its location

Differential TDR Measurements

- Requires Two incident step pulses to stimulate the DUT
- TDR module has a unique capability to generate a *positive* and a *negative* step pulse
- The positive and negative pulses from the two TDR channels are *turned ON simultaneously for a TRUE differential measurement*
- De-skew* between the two pulses is performed as part of the Normalization Procedure

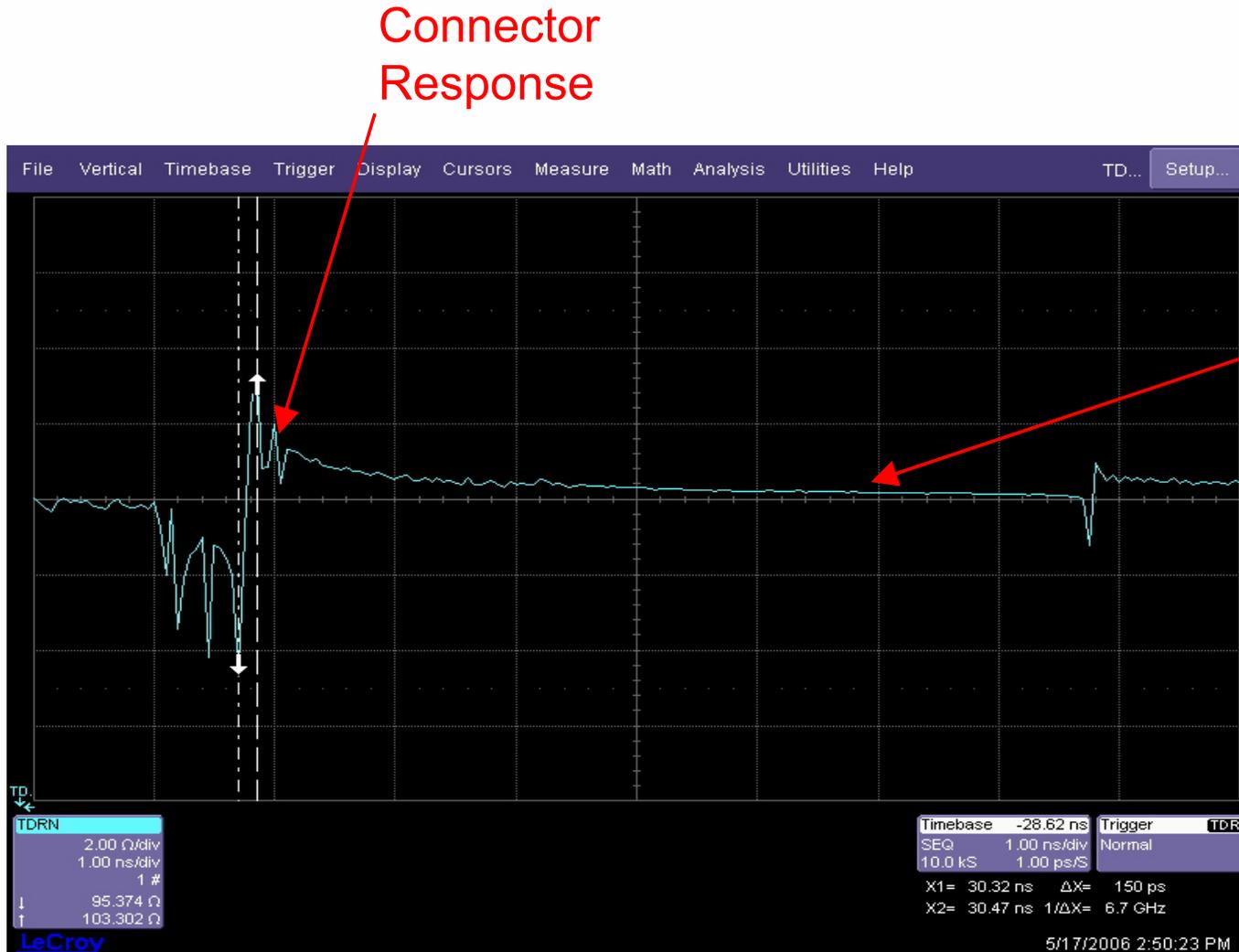


Differential devices - under differential and common-mode stimulus



Differential TDR meas. Set-up

Differential Measurement Example



100 Ω Reference

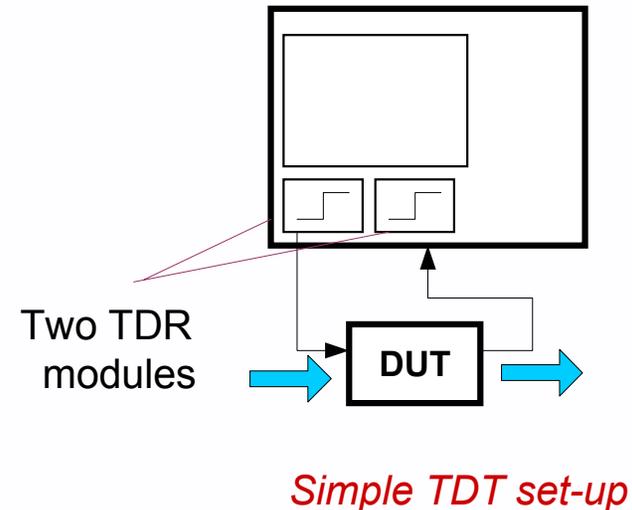
Differential TDR measurement

Differential Measurement Highlights

- *True Differential* input signal to the Device Under Test (DUT) by turning ON the two TDR channels simultaneously
- De-skew procedure removes
 - Sampling skew between modules
 - Skew between the Positive and Negative Step Pulses
 - Additional skew from external cables/connectors connected to the two channels
- Normalization procedure removes effect of external cables and adapters used in the set up for accurate differential measurements.

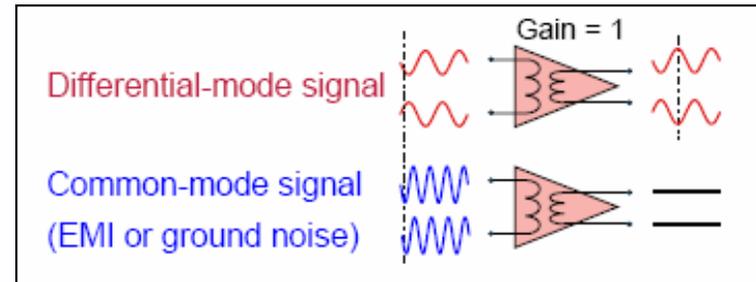
What is TDT?

- **(TDT) Time Domain Transmission**
Measure of the effects on a signal transmitted through a Device Under Test (DUT)
- Used to measure *Insertion (Transmission) Loss S_{21}* through a system.
- TDT response converted to *frequency* response (also known as Transmission *S-parameter*) can help determine the effects on an eye diagram -
- Requires two TDR Channels - one to generate the step and other to sample

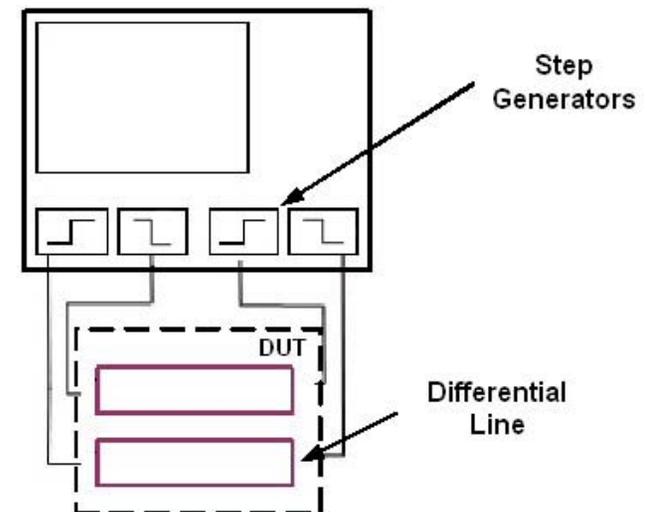


Differential TDT

- **Differential TDT** - Measure of the effects on a *differential signal* transmitted through a Device Under Test (DUT)
- Used to measure *Differential Insertion (Transmission) Loss* through a system.
- Requires ***four*** TDR Channels - two to generate the step and other two to sample



Differential devices – under differential and common-mode stimulus



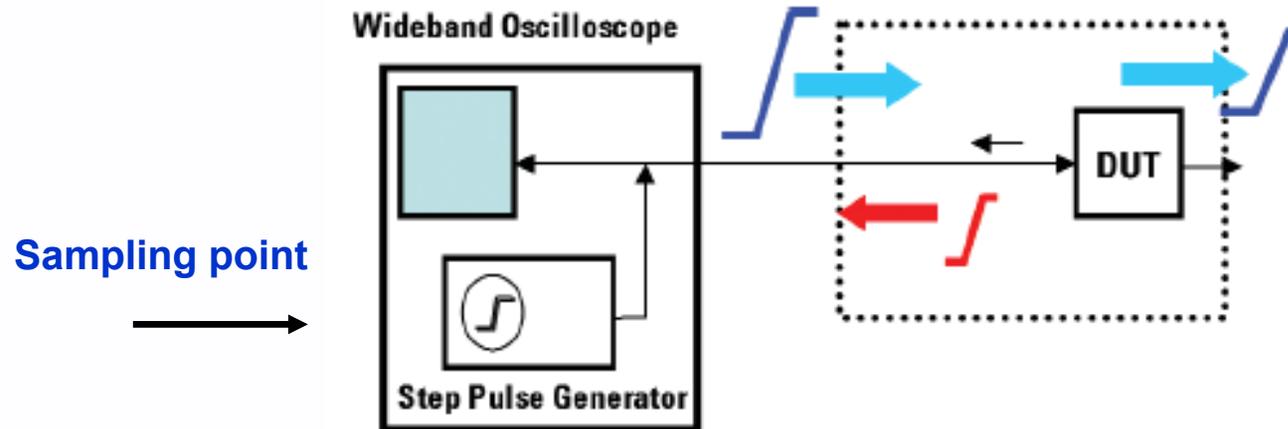
Differential TDT meas. Set-up

TDR Measurement Basics

What Can You Extract from TDR/TDT Measurements?

- ***TDR Reflections*** help analyze the discontinuities in high speed circuits including the type (*short circuit, open circuit, capacitive etc.*)
- ***Location of discontinuity, Impedance, Crosstalk*** is calculated from the same measurement
 - Impedance tolerance specs. are now part of all digital transmission systems - USB 2.0, PCI-Express, FBDIMM etc.
- The acquisition of ***calibrated frequency dependent network parameters (S-parameters)***

TDR Impedance calculation



- The measured signal by the scope is

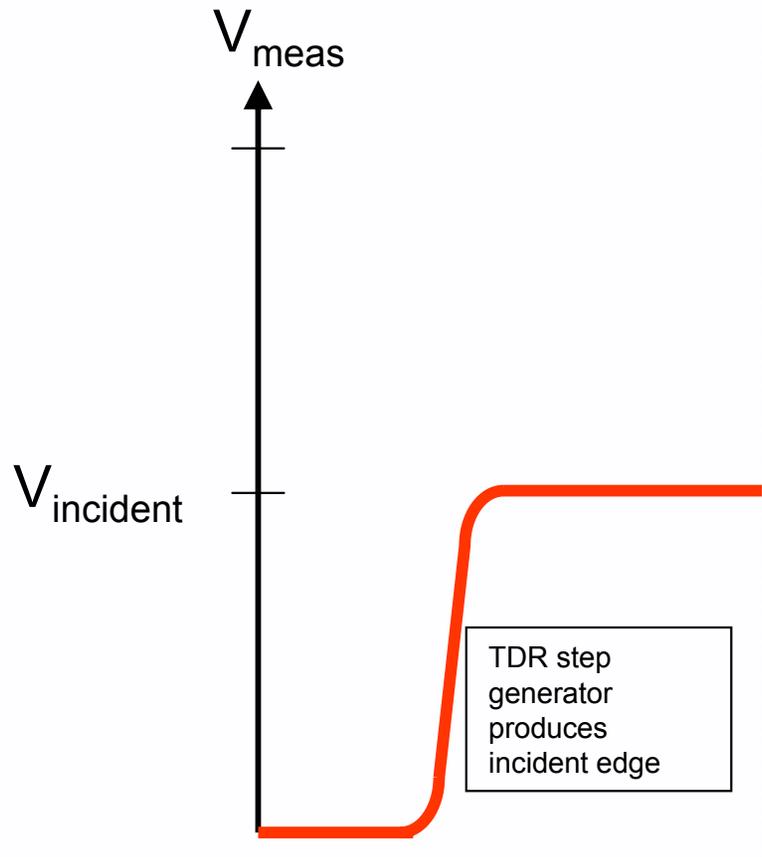
$$V_{\text{measured}} = V_{\text{incident}} + V_{\text{reflected}}$$

$$\text{Reflection, } \rho = \frac{V_{\text{reflected}}}{V_{\text{incident}}} = \frac{Z_L - Z_0}{Z_L + Z_0}$$

$$V_{\text{reflected}} = V_{\text{incident}} \left(\frac{Z_L - Z_0}{Z_L + Z_0} \right)$$

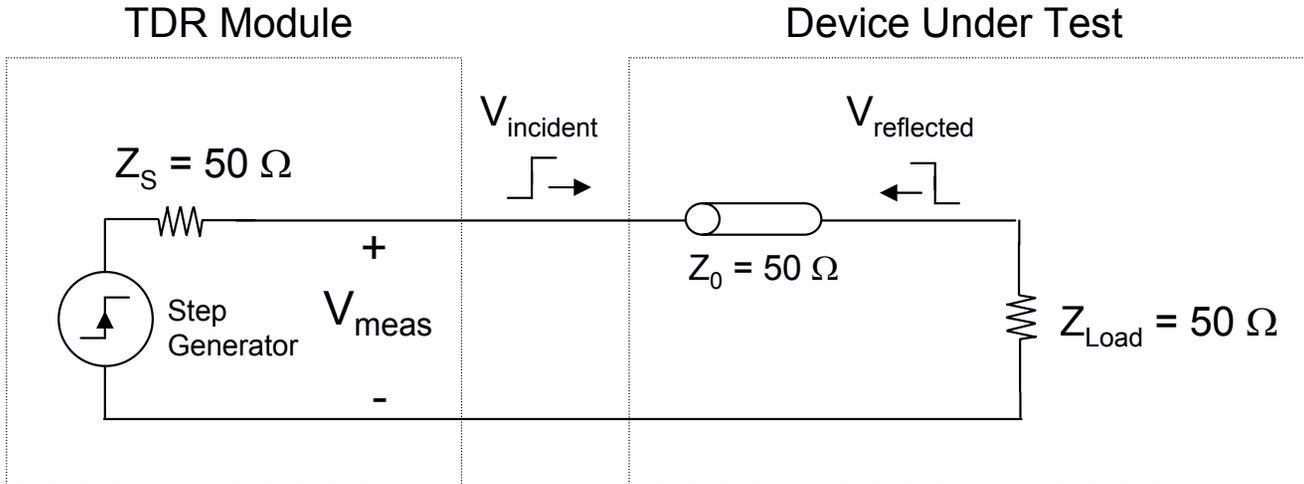
Basic TDR measurements

An incident step from the TDR (TDR module) is used as stimulus



Depiction of an incident pulse from the TDR

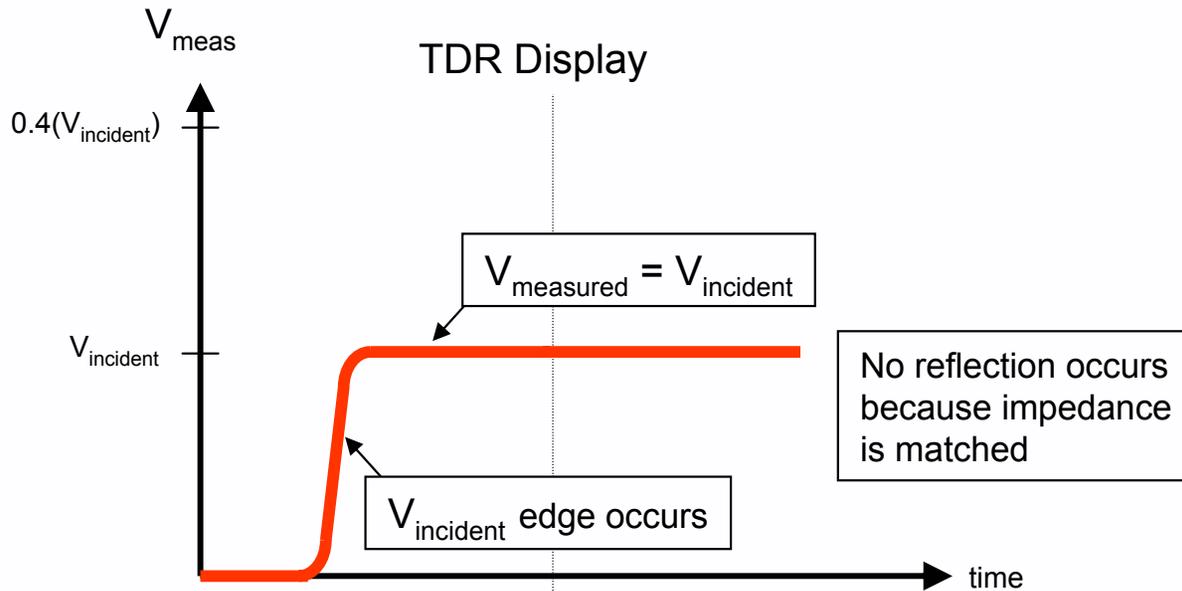
TDR Example: Matched Impedance ($Z_L = 50 \text{ ohm load}$)



Solving for $V_{\text{reflected}}$:

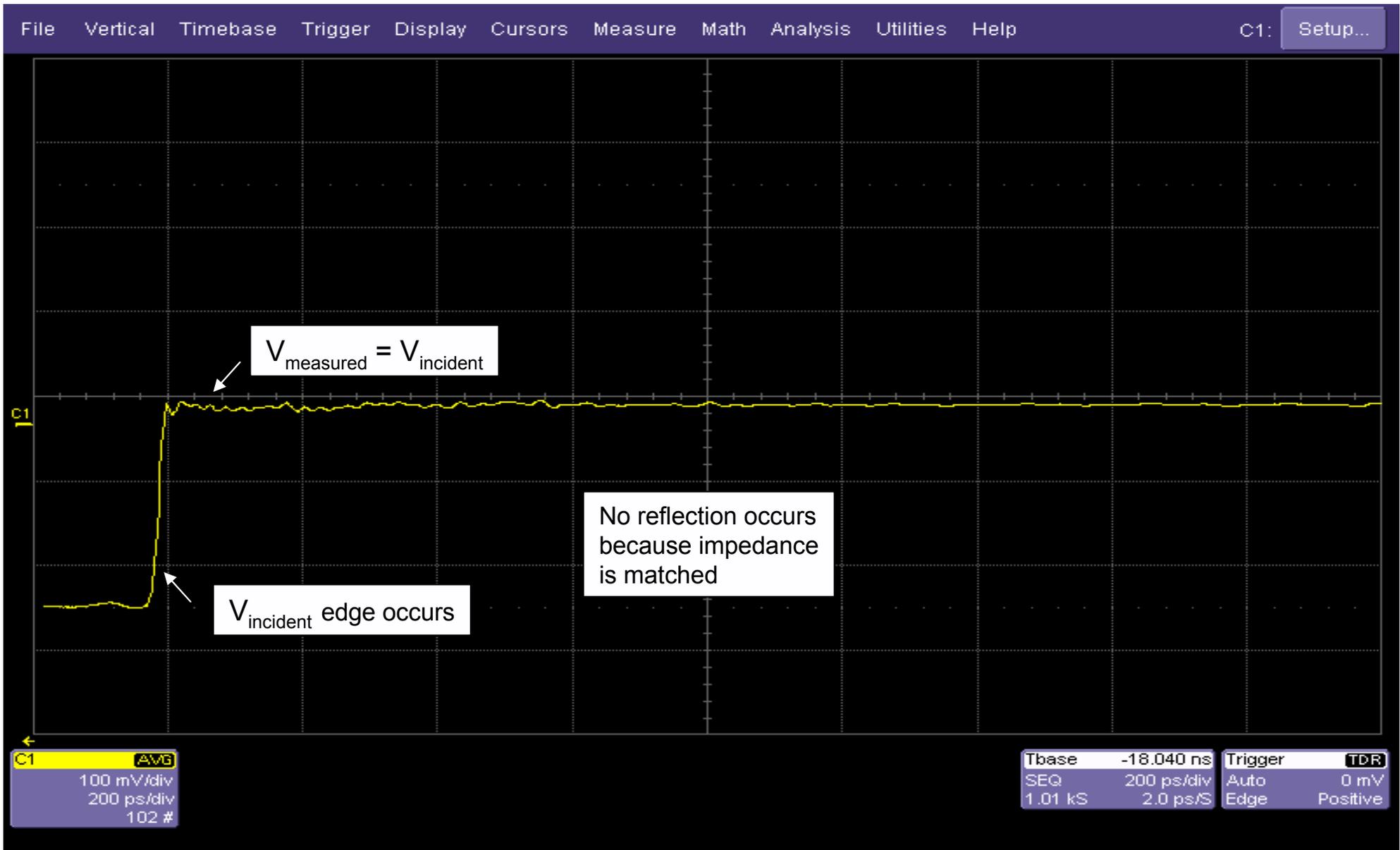
$$V_{\text{reflected}} = V_{\text{incident}} \left(\frac{Z_{\text{Load}} - Z_0}{Z_{\text{Load}} + Z_0} \right)$$

$$Z_{\text{Load}} = Z_0, \quad \ominus V_{\text{reflected}} = 0$$

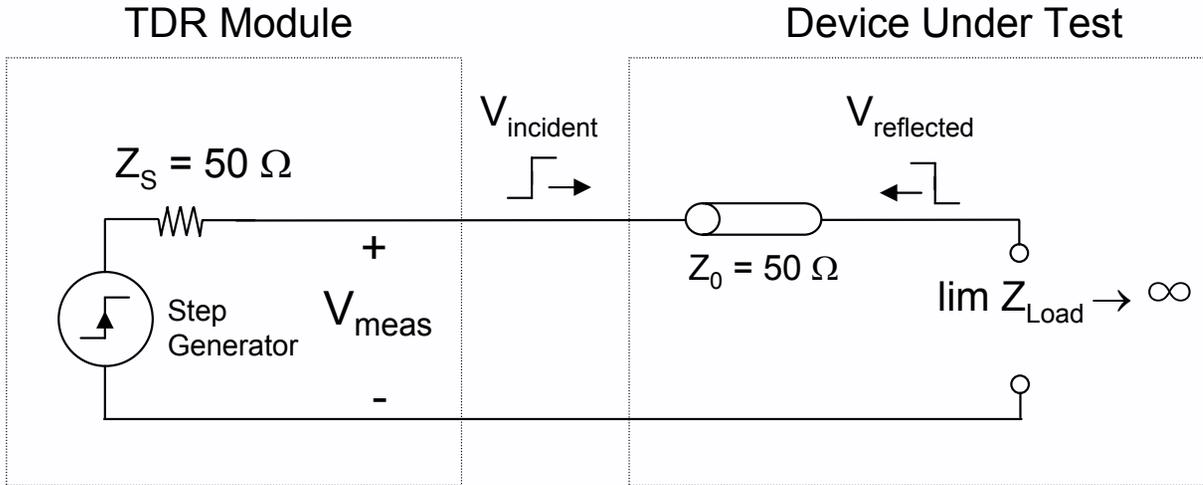


The most important design technique to control signal quality:
“Keep the impedance the signal sees constant all along the net”

TDR Example: Matched Impedance



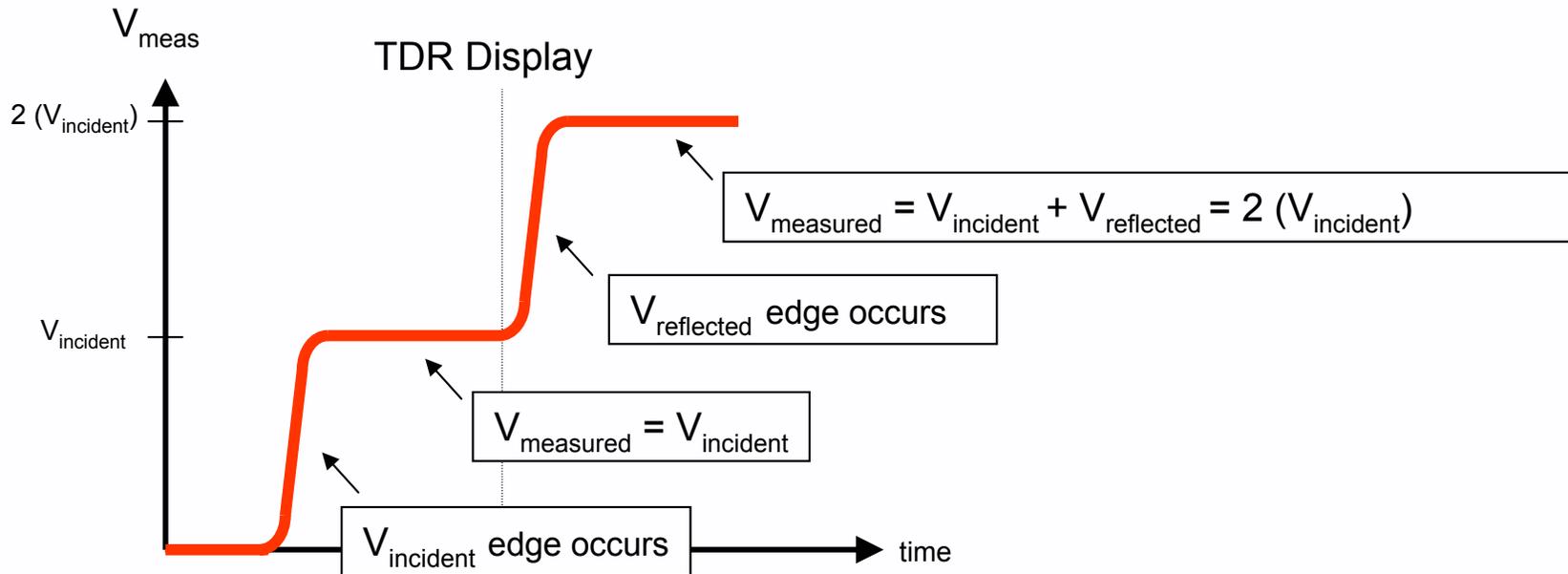
TDR Example: Open Circuit ($Z_{Load} \rightarrow \infty$)



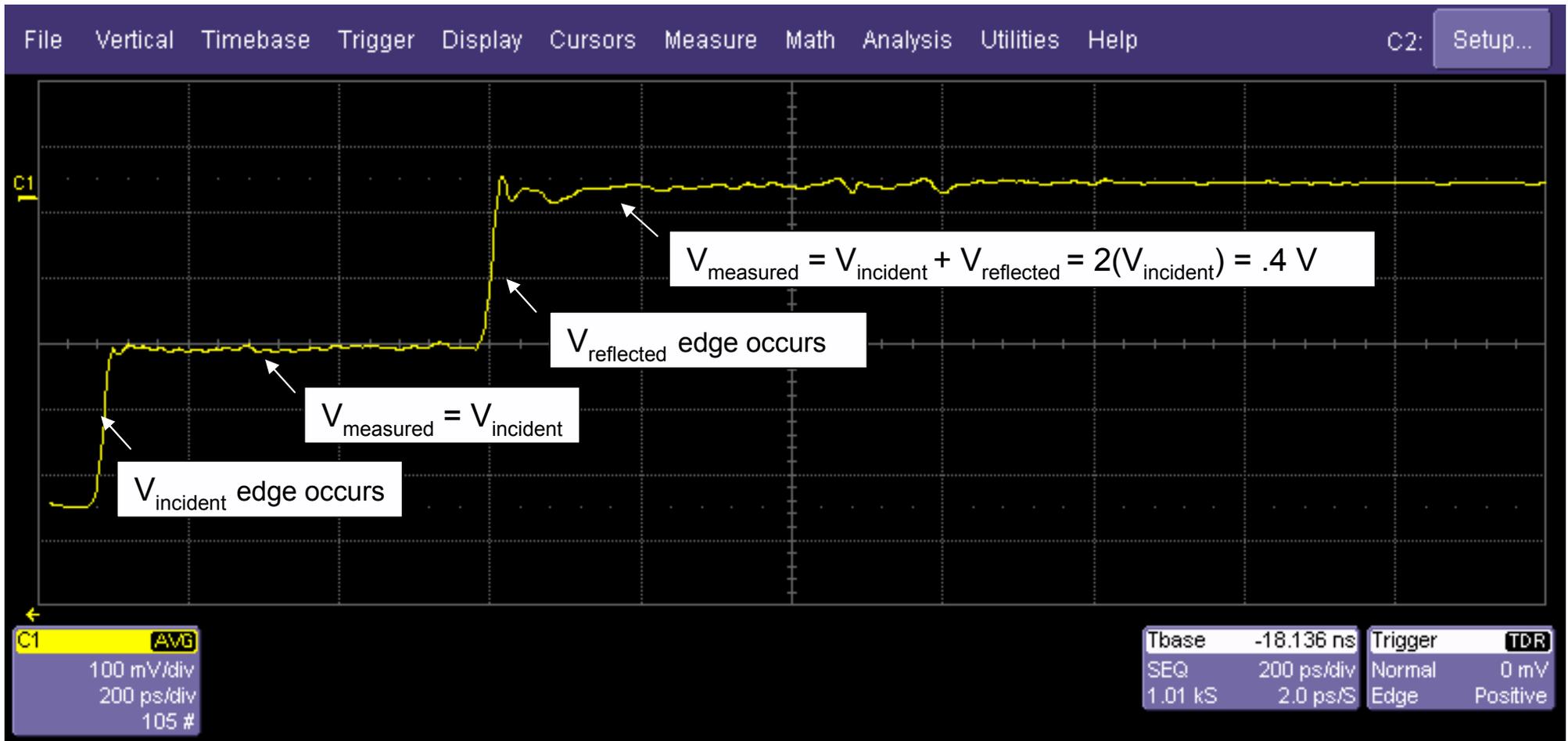
Solving for $V_{reflected}$:

$$V_{reflected} = V_{incident} \left(\frac{Z_{Load} - Z_0}{Z_{Load} + Z_0} \right)$$

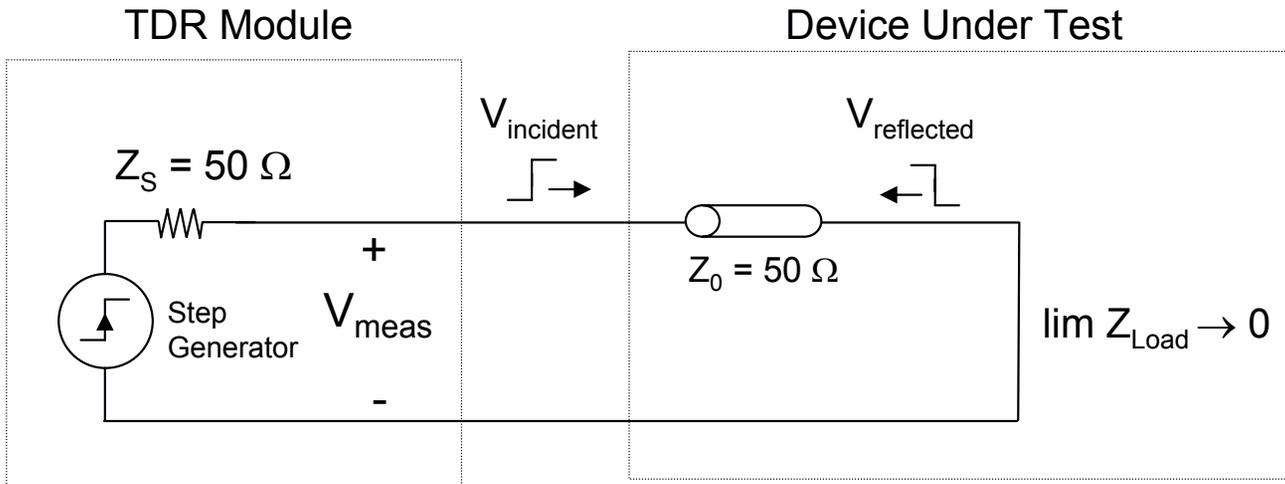
$$\lim Z_{Load} \rightarrow \infty, \Omega V_{reflected} = V_{incident}$$



TDR Example: Open in the air



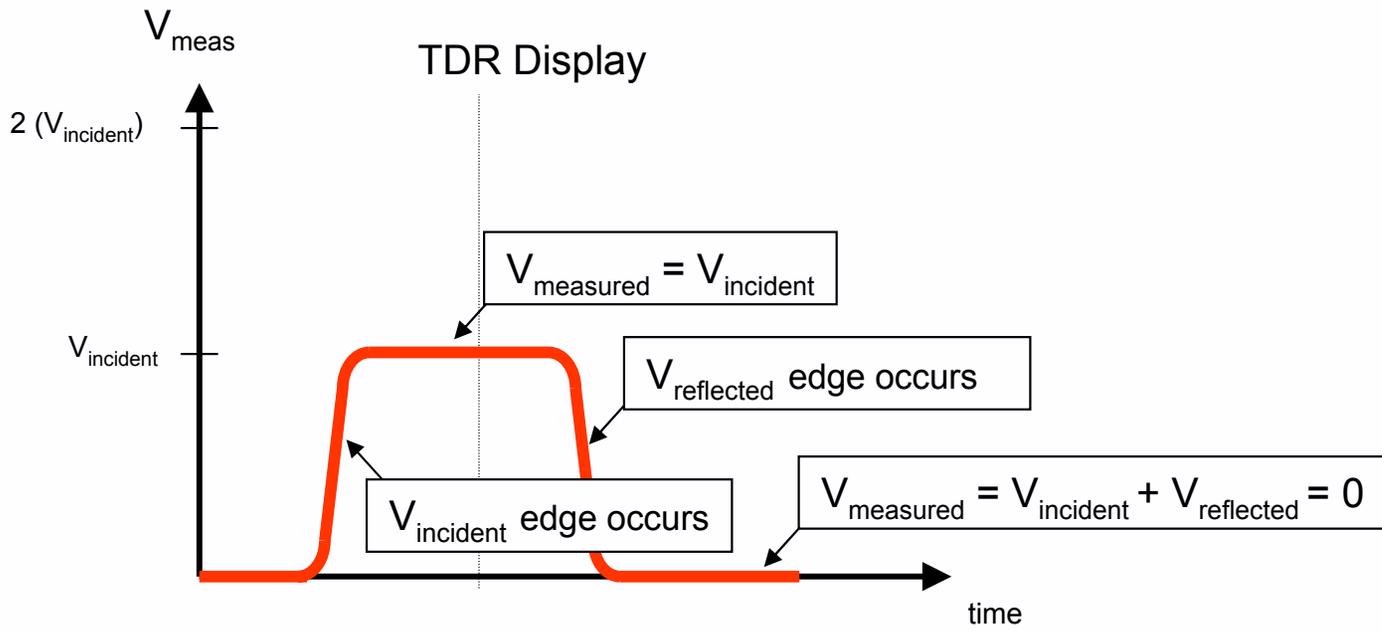
TDR Example: Short Circuit ($Z_{Load} = 0,)$



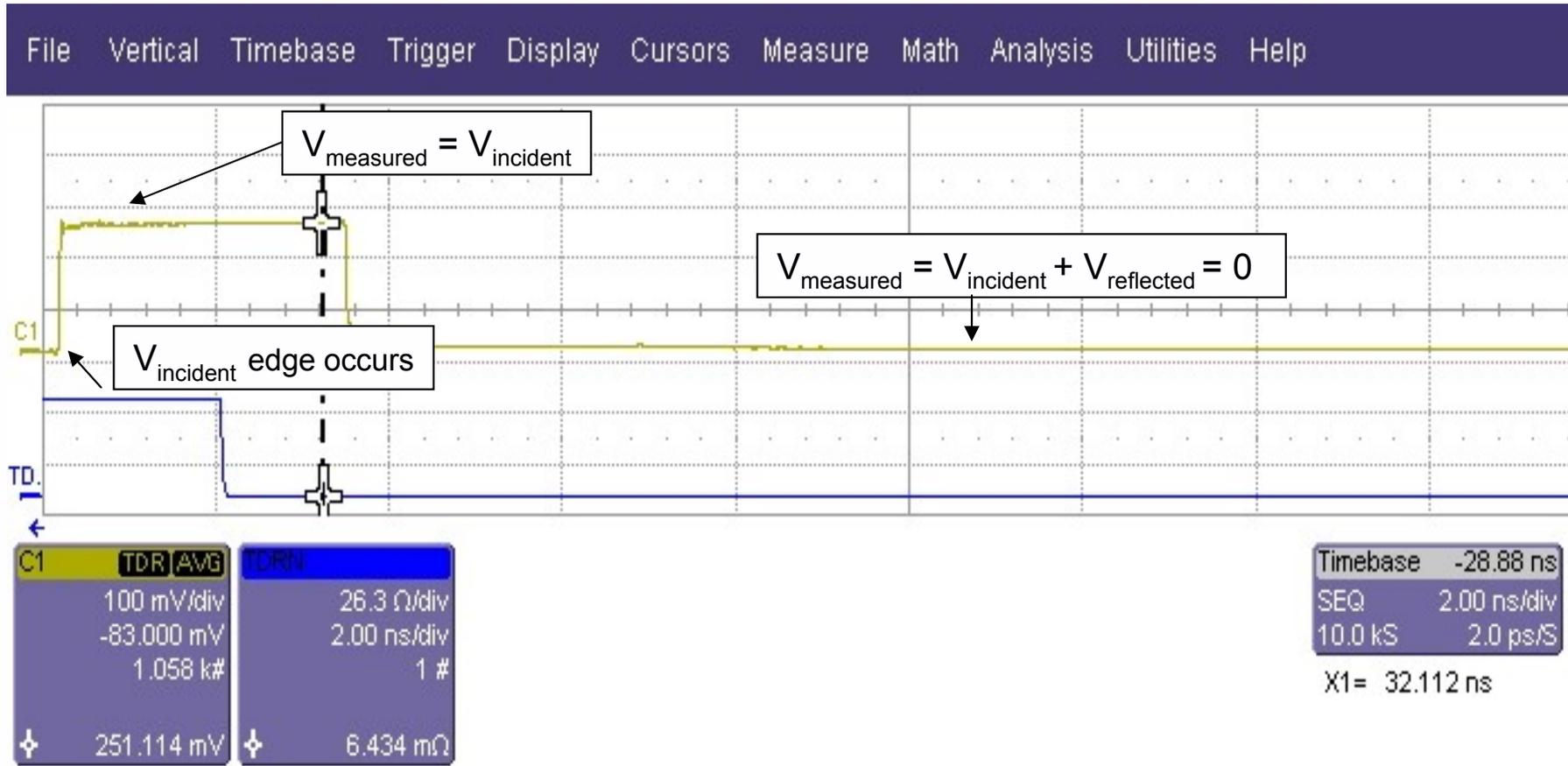
Solving for $V_{reflected}$:

$$V_{reflected} = V_{incident} \left(\frac{Z_{Load} - Z_0}{Z_{Load} + Z_0} \right)$$

$$\lim_{Z_{Load} \rightarrow \infty} V_{reflected} = -V_{incident}$$

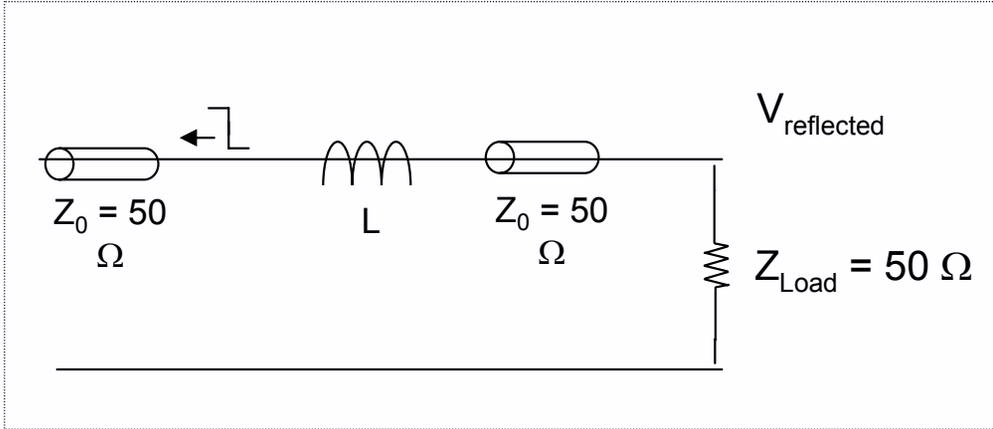


TDR Measurement: Short ($Z_L = 0$)

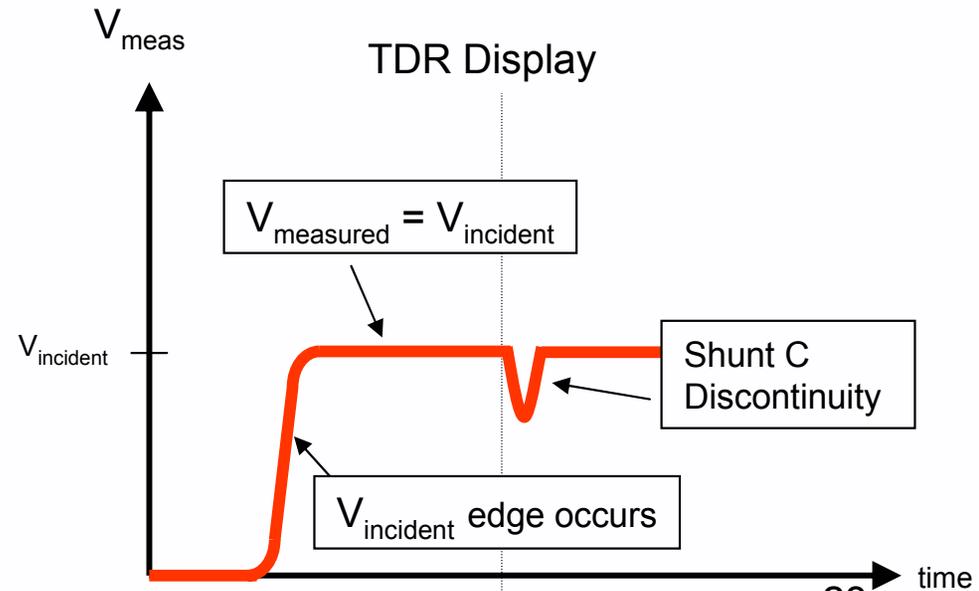
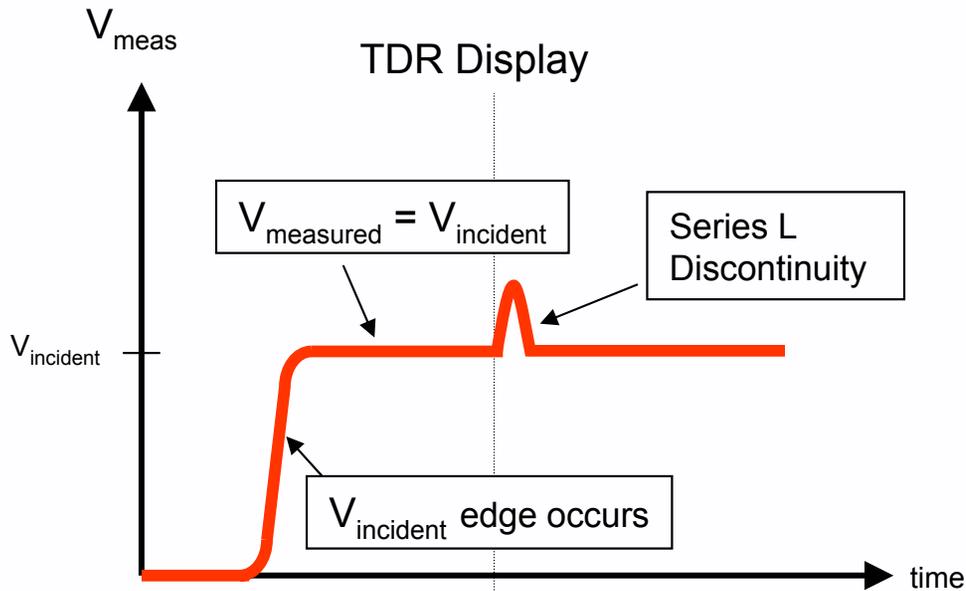
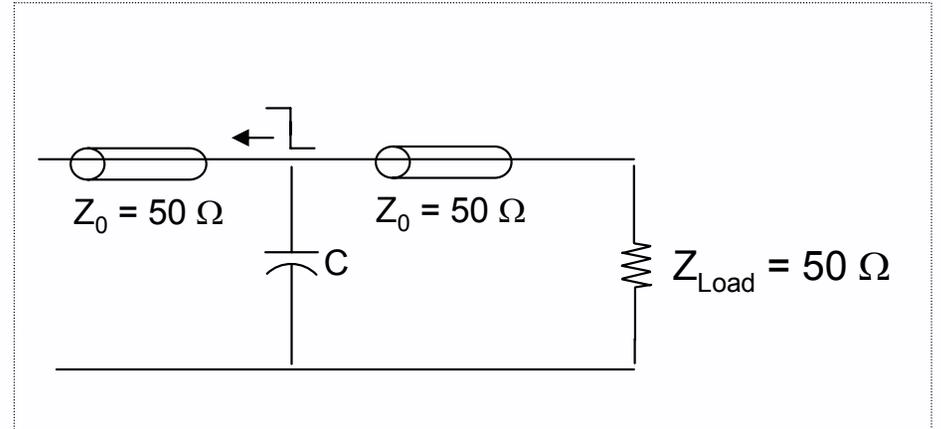


TDR Example: Reactive Elements - L and C Discontinuity

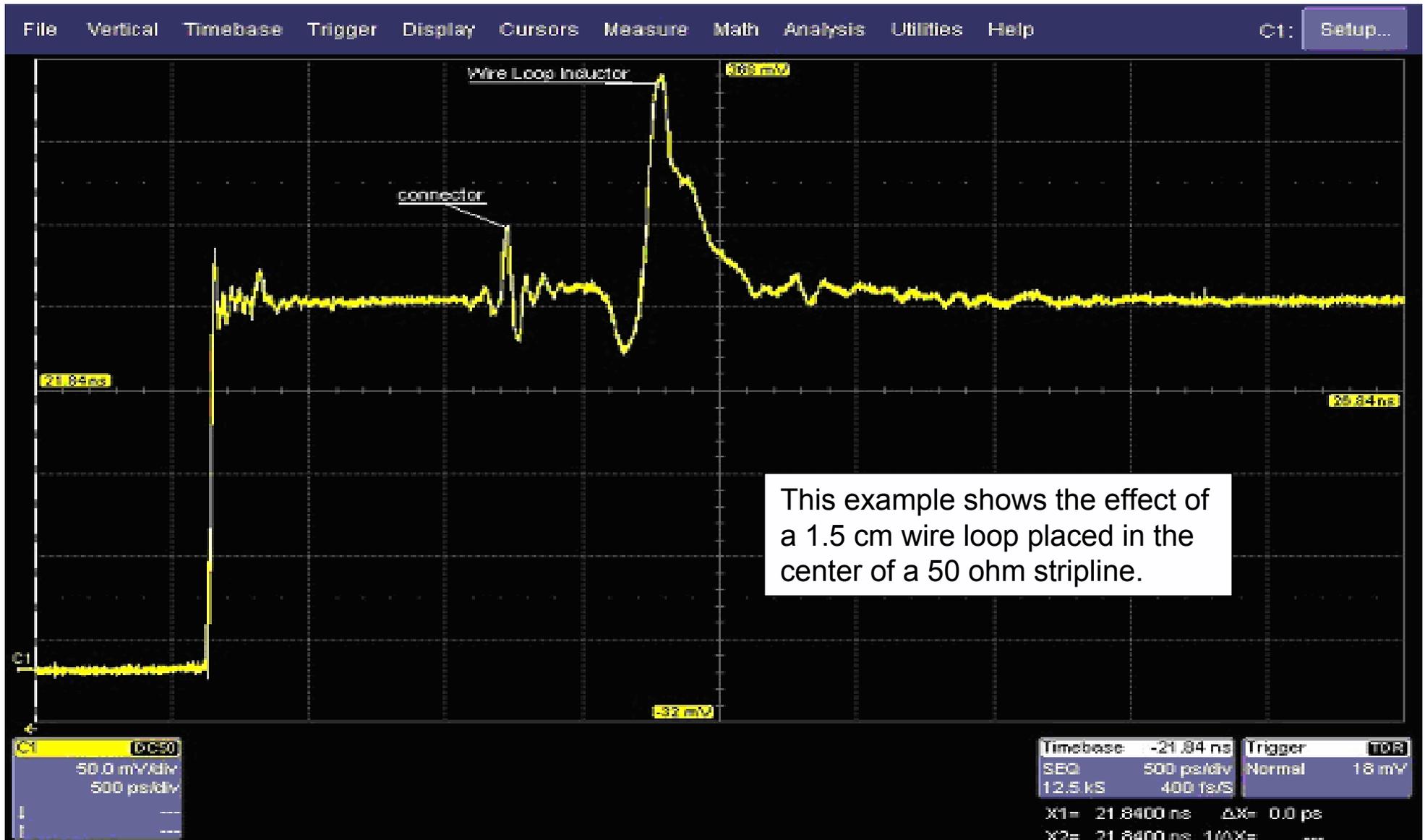
Device Under Test



Device Under Test

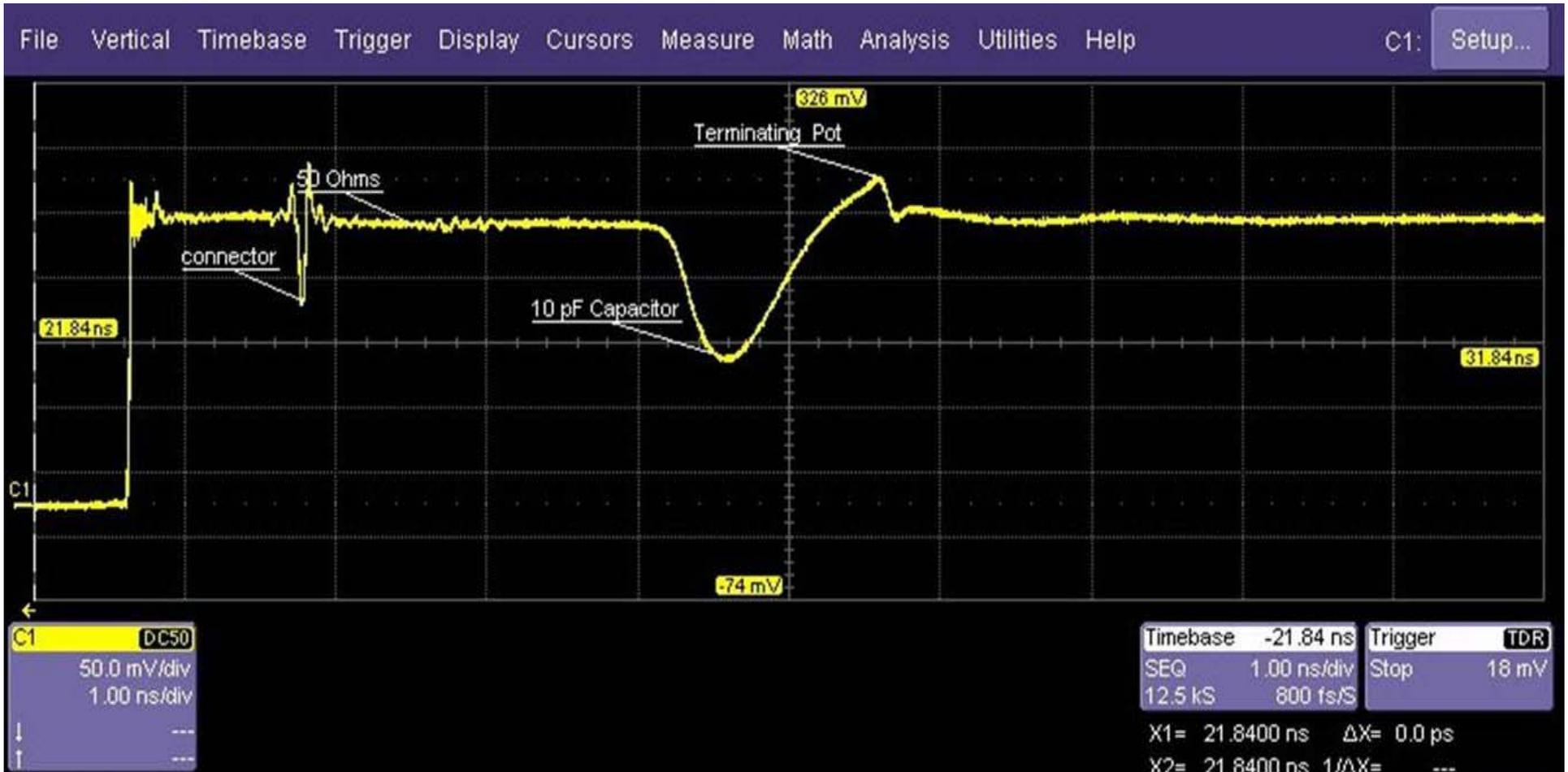


TDR Example: Series L Discontinuity



This example shows the effect of a 1.5 cm wire loop placed in the center of a 50 ohm stripline.

TDR Example: Shunt C Discontinuity

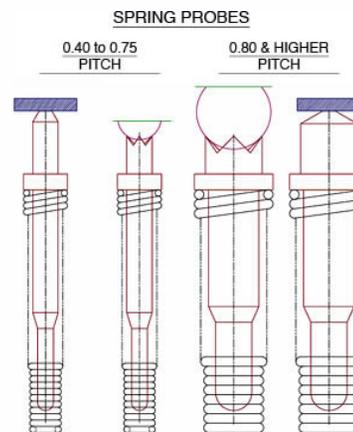
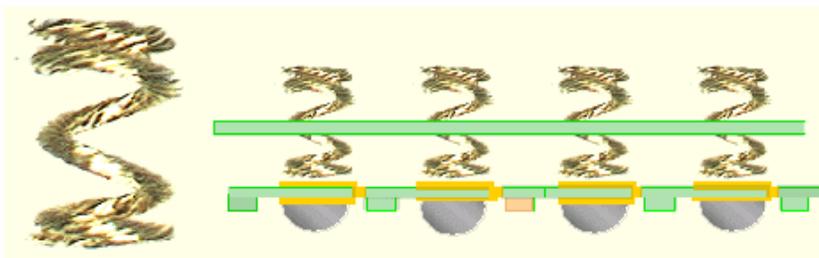
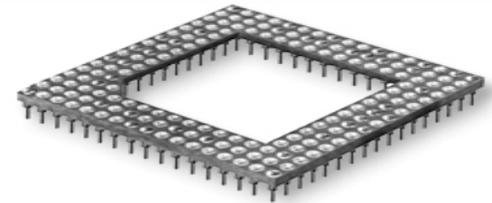
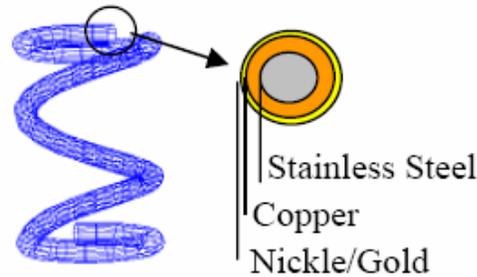
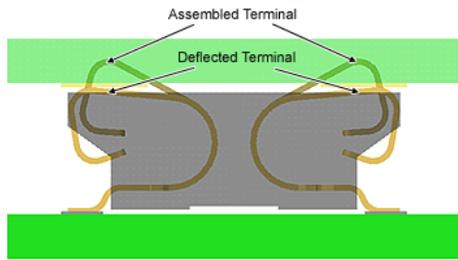
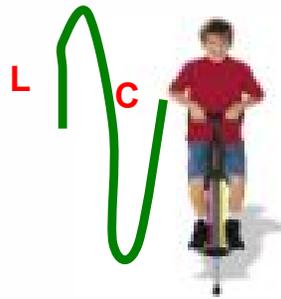
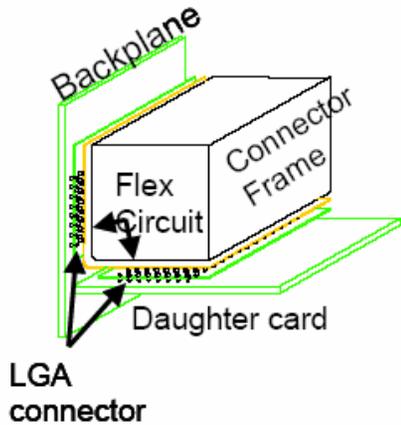


Source of L/C Discontinuities

Compression Connectors

Socket interposers Pad to Pad compliance is achieved through Spring Compression Contacts

- Induces L/C discontinuities
- Degrades interconnect system Return loss In Gbit data rates



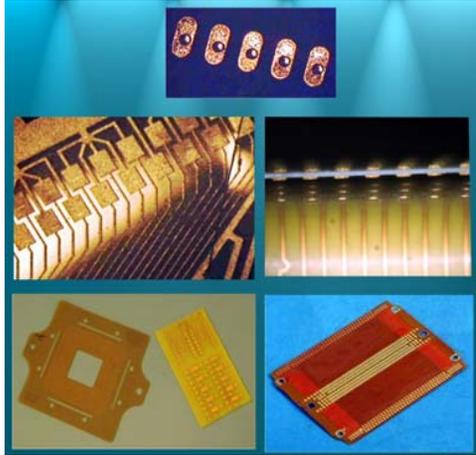
U.S. PATENT #6844749



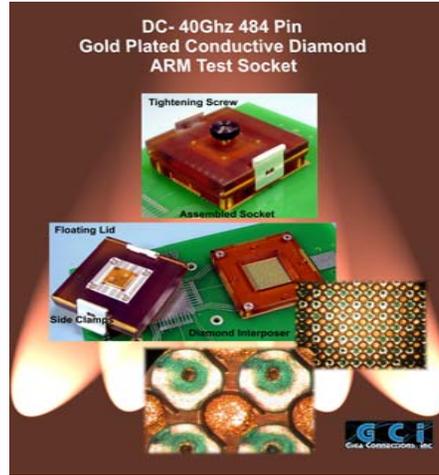
Compression Connectors are found in:

- @ Board to board connectors
- @ Test sockets
- @ Package-to-Board Connectors

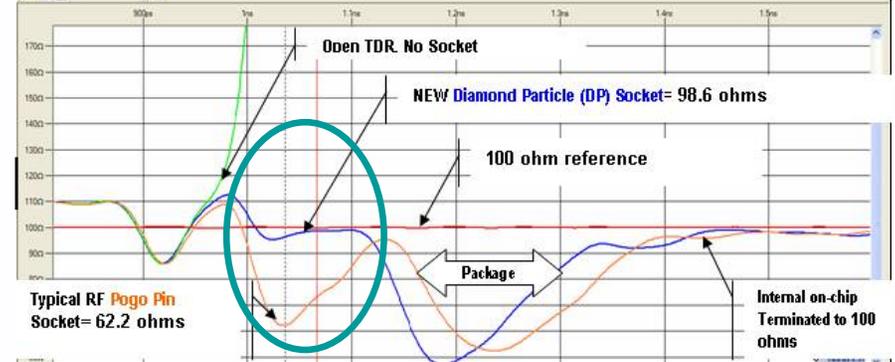
New 40Ghz ~~L/C~~ Interconnect Technology Gold Plated Conductive Diamond Interposer



Board to Board Interconnects

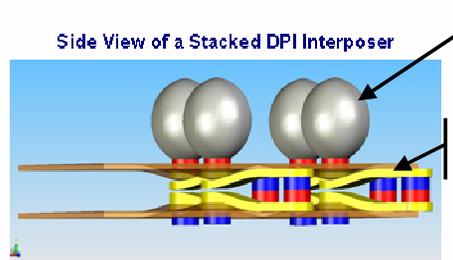
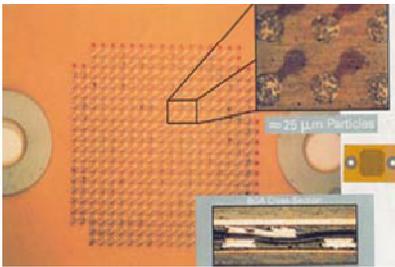


Pin-less RF Test Sockets



The (DP) socket is a better match for the PCB and the transceiver IC onboard terminator. Most significantly, the conductive diamond particles show no capacitive or inductive reactance. With a uniform impedance profile, the (DP) socket decreased Return Loss by 3dB.

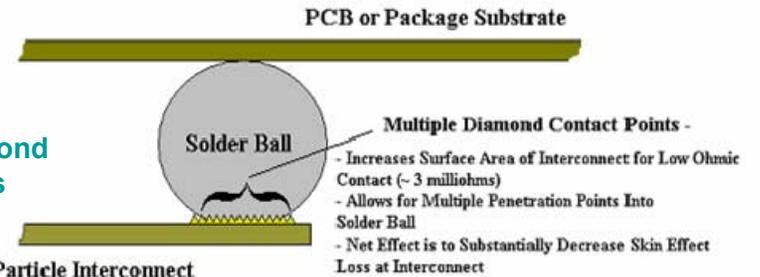
No Measurable L/C to 40Ghz



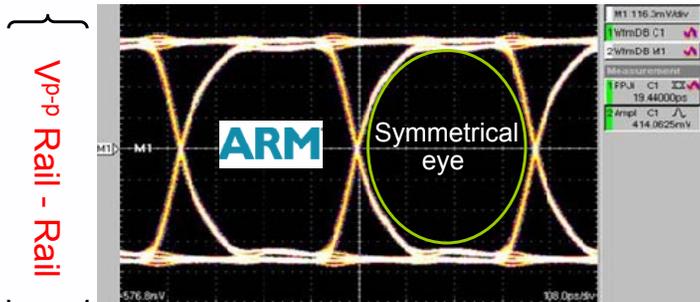
BGA

Stacked Diamond Interposers

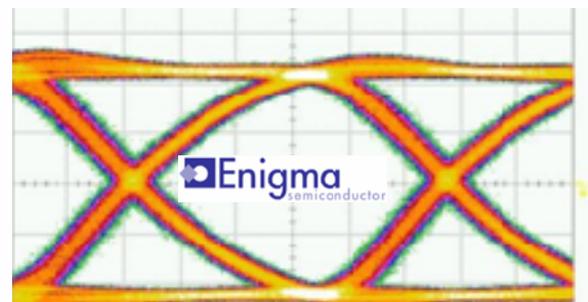
Diamond Particle Interconnect



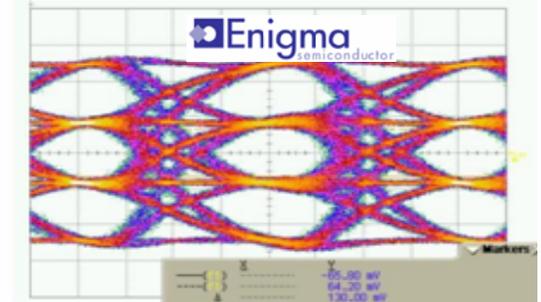
BGA Compliance using stacked off-set interposers



6Gbit PCI Express



6Gbit Custom IP Device



10 Gbit 4PAM

Gold Plated Conductive Diamond Interposer Specifications

Interposer Specifications

Signal Integrity

- > 40 GHZ bandwidth
- No measurable inductance, or capacitance
- ~1 % increase in impedance measured at 10gbits data rate

Interposer Interconnect Technology

- 10-20 um conductive diamonds
- 10 micro-inches gold plated
- Contact point size is 10um; 10-mil pitch can be achieved.
- RoHs compliance - lead free

Pin Count

- >2000

Maintenance

- Surface can be refurbished
- Clean with ultrasound bath

Reliability

- 100's of thousands of insertions without measurable degradation in signal integrity

Force

- 10-15 Grams force per pad required to achieve electrical continuity

Current Handling

- ~15 amps per 10 mil diameter pad

Temperature Range

- -60C to 200C for Kapton film Interposer

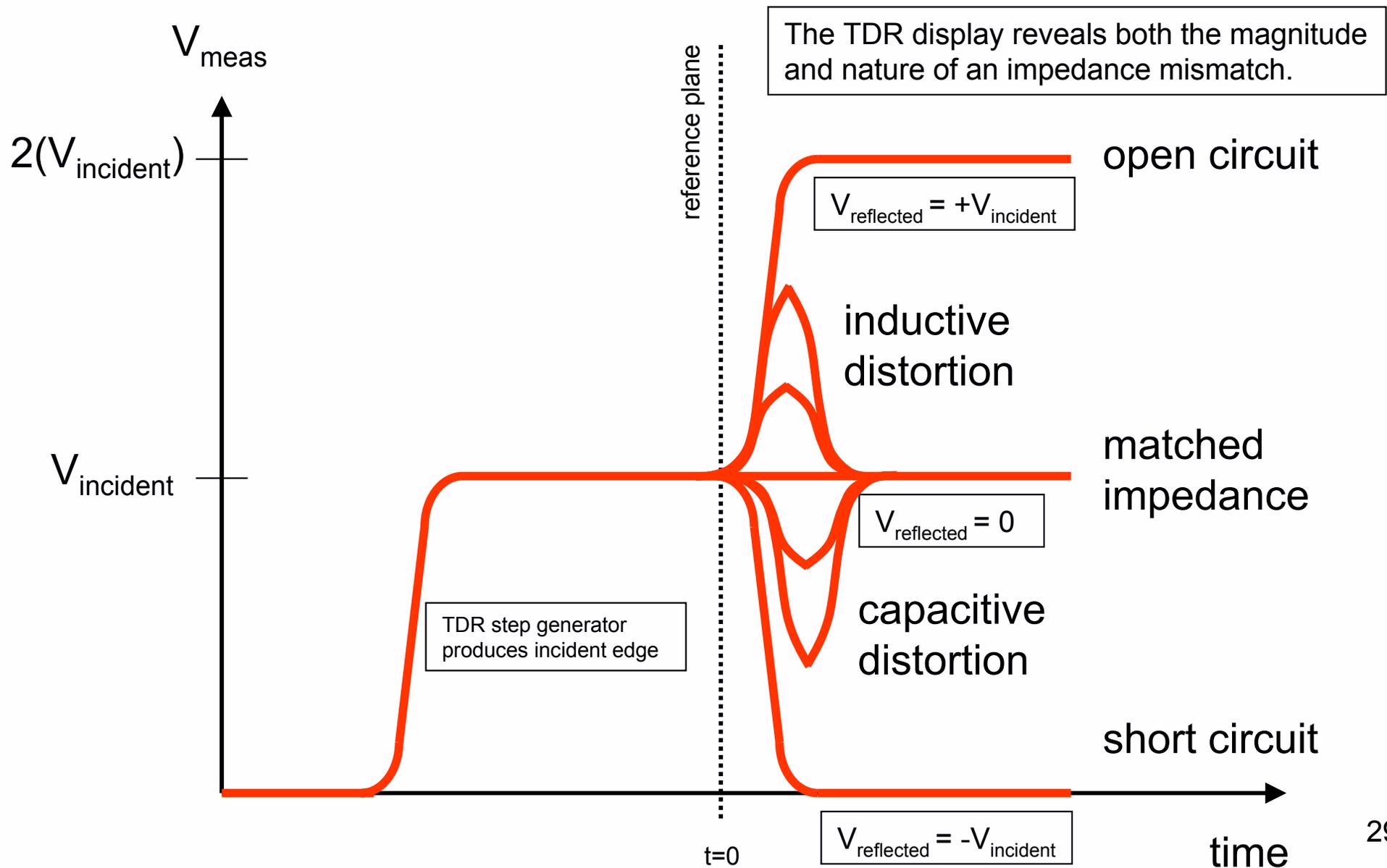
[Available from](#)

Giga Connection, inc.

www.gigaconnections.com



Summary: Interpreting the TDR Display



TDR measurements

Locating Discontinuity (open/short) Fault

Q. How do I know where the fault is in the system??

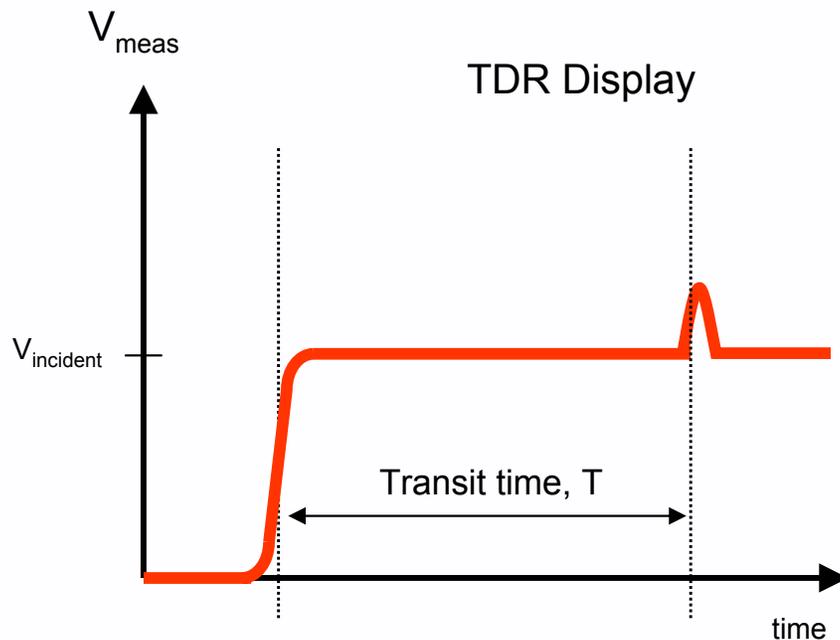
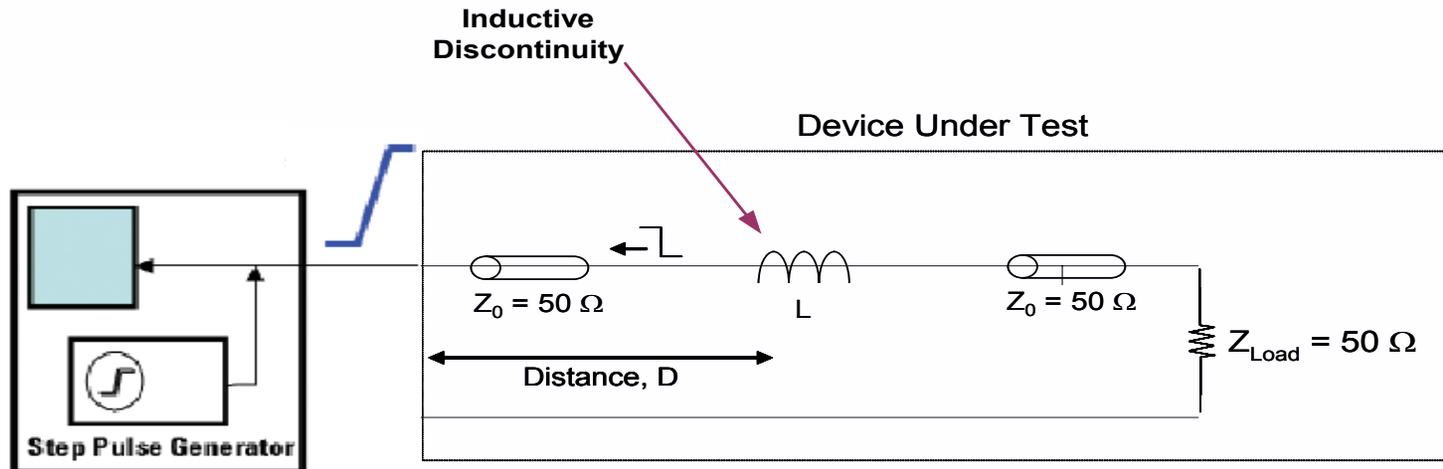
- *Scope measurement yields location of the faults in **time***
- ***Time** can be converted to **distance** by entering the propagation velocity (v_p) or dielectric constant (ϵ_R) of the medium*

NOTE: *Dielectric constant of a medium may vary with distance, hence adding in a constant value for (ϵ_R) may not always yield the accurate distance*

Typical dielectric constant of FR4 = 4.2

The two main constituents of FR4 are glass with an E_r of around 6 and resin with a dielectric constant of around 3.

TDR Example: Determining Fault Location



Physical distance to fault location can be determined by:

$$D = 0.5 * (T) * (v_p)$$

D = physical distance to fault location
 T = transit time from monitoring point to mismatch
 v_p = velocity of propagation (material property)

Enhancing Accuracy of TDR measurements

- Use of low loss cables, adapters
- Use of extender cables to avoid losses due to using long cables
- Long memory length for better impedance resolution when measuring long interconnects like backplanes, Cables, etc.

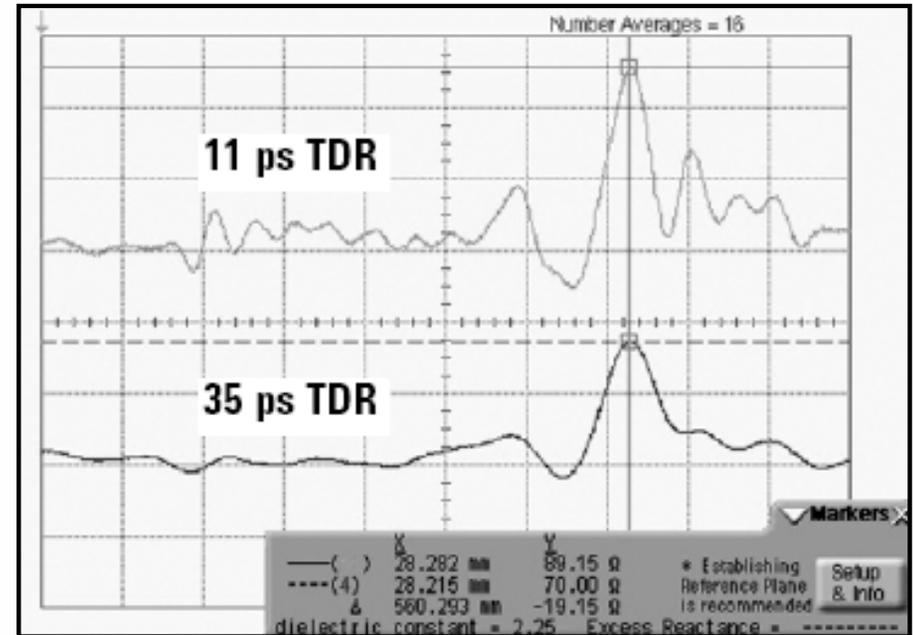
TDR Rise Time considerations

- 1) **FAST Rise time** results in better measurement resolution
- **18ps** (typical) Rise time for TDR
- Resolution $\approx \frac{1}{2}$ (Step rise time)

$$\text{WaveExpert} = \frac{1}{2} (18 \text{ ps})$$

$$\approx 9 \text{ ps (1.25 mm on PCB)}$$

Closely spaced discontinuities can be measured with higher accuracy



Measurement of a connector assembly using pulses with different rise times – Faster rise time pulse yields better measurement resolution

Calibration (Normalization)

TDR Calibration (Normalization)

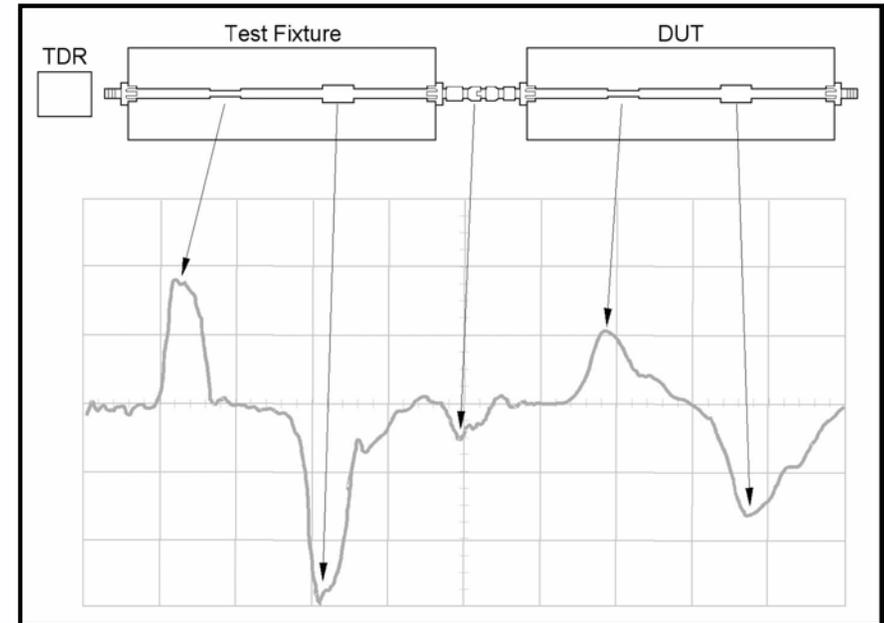
Why is it required??

- We don't live in a PERFECT WORLD!

Errors in TDR measurement:

Typical TDR set-up will include:

- Cables, Adapters, etc
- Test fixture (if DUT is on a PCB)



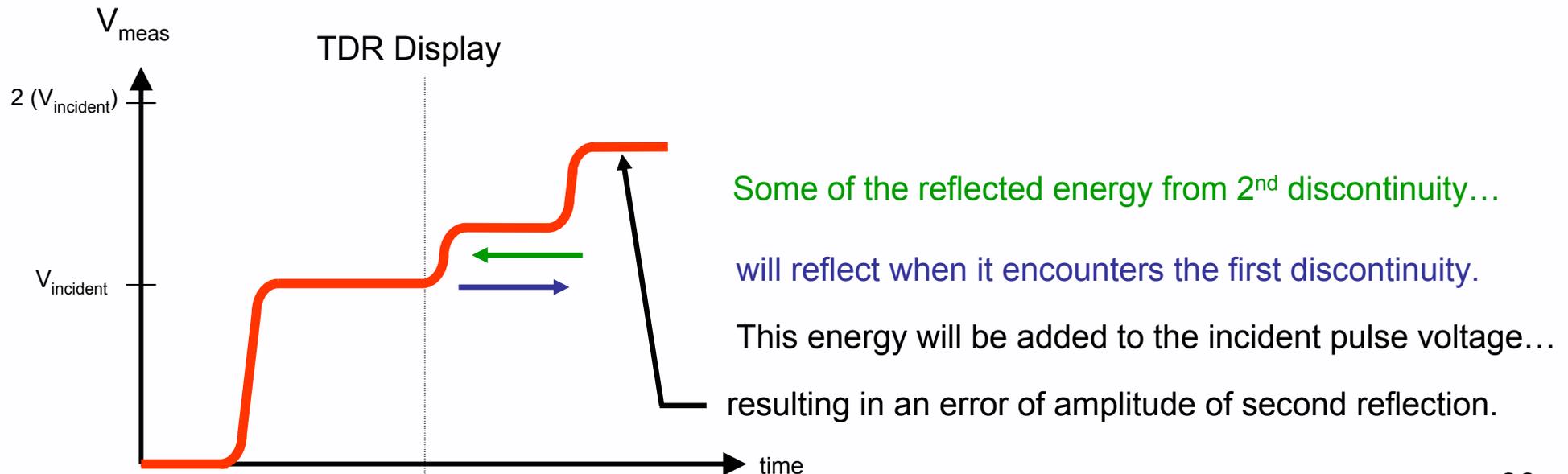
Typical TDR measurement set-up consisting of cables and test fixture

TDR measurement of the DUT will yield composite reflections from all the components present in the test path

$$V_{refl} = V_{refl}(\text{Test fixture}) + V_{refl}(\text{conn.}) + V_{refl}(\text{adapters}) + V_{refl}(\text{DUT})$$

Back Reflections

- TDR techniques have some limitations.
- Most users are aware of these:
 - Major discontinuities cause "back reflections", resulting in reflected energy of the return reflection from additional discontinuities.
 - Thus, the accuracy of impedance measurements after the first major discontinuity is reduced.



TDR calibration (contd.)

- These errors can be corrected by performing a calibration with known standards.
- This adjusts the reference plane such that accurate time/distance measurements of the DUT are possible

Other sources of errors in TDR measurements:

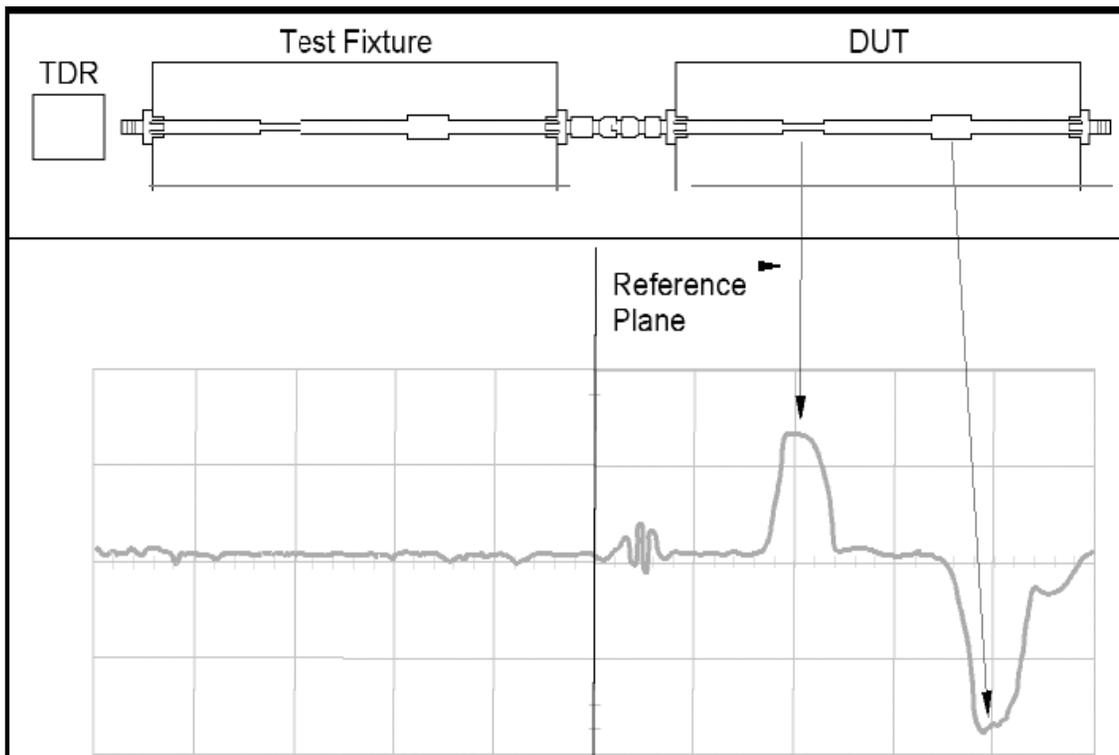
- Accuracy of the step from the TDR module
- Oscilloscope response

Errors that cannot be calibrated:

- Temperature drifts
- Connector repeatability

TDR Calibration (contd.)

After calibration – effects of test fixture and connectors are removed from the response



Reference Plane:

The input reference point to the DUT for TDR meas.

Calibration/Normalization features

TDR Module Calibration

- Amplitude Variation correction

Reference Plane Calibration/Normalization

- Removal of Test Fixture/Cables/Adapters from measurement
- Pulse De-skew capability for differential measurements
- Coaxial 3.5 mm Calibration Kit

TDR Calibration Methods

Short Load (SL)

- Uses Two known standards to calibrate the oscilloscope for TDR measurements
 - Short standard (Impedance $Z_L = 0$ at input reference of DUT)
 - Load (or Match - an ideal 50Ω impedance standard)
- Most commonly used calibration for TDR measurements

TDR Calibration Methods

Open Short Load (OSL)

- Uses Three known standards to calibrate the oscilloscope for TDR measurements
 - Open standard (Open circuit at input reference of DUT)
 - Short standard (Impedance $Z_L = 0$ at input reference of DUT)
 - Load (or Match - an ideal **50 Ω** impedance standard)
- More accurate calibration technique than SL method; can characterize reflections better
- Coaxial Calibration kit from Maury Microwave

Differential TDR Calibrations

- Require simultaneous drive to the DUT (min. two TDRs)
 - one positive step and one negative step
- Any skew between the step pulses affects measurement accuracy
 - Cables connecting to DUT should be of similar lengths for de-skew procedure to work
- Reference Impedance is **100 ohms**
 - Still calibrated by using 50 ohm standards on each ports
- Normalized (Calibrated) trace displays a differential response

S-parameters

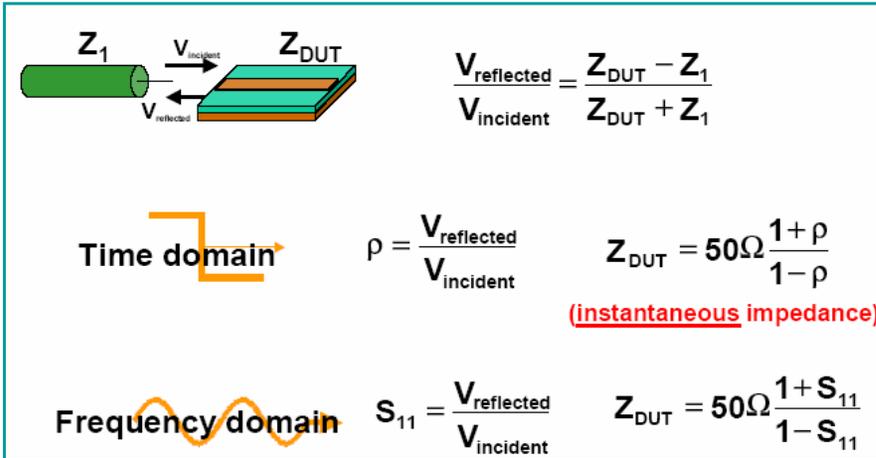
S-parameters (Scattering parameters) are ratios of power that represent the frequency performance of a device. Each S parameter is the ratio of the sine wave coming out to the sine wave going in the DUT.

S₂₁ for example, would be the ratio of the sine wave voltage coming out of port 2 to the sine wave voltage that goes into port 1.

Typical S-parameter terms include: Return Loss, Insertion Loss, and Crosstalk.

T_{DD11}	differential impedance profile
S_{DD21}	Signal quality of differential signal, time delay of differential signal
T_{CD11}	Conversion of differential signal to common signal in reflection (emissions)
S_{DC21}	Conversion of common signal to differential signal in transmission (susceptibility)
T_{CC11}	Common impedance profile
S_{CC21}	Signal quality of the common signal, time delay of common signal

Impedance In the Time and Frequency Domains



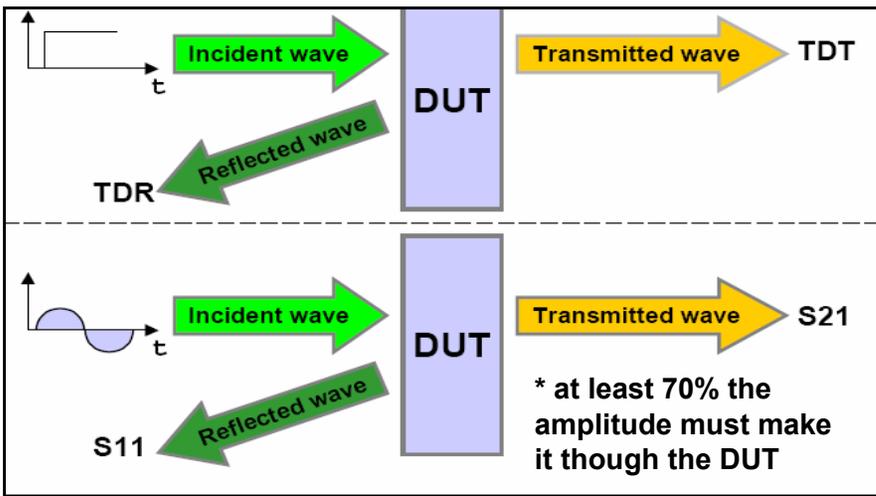
Time and Freq. Measurements

Time Domain

- **TDR** (S11) Impedance vs time discontinuities
- **TDT** (S21) Dielectric, Skin Effect loss
 - Noticed as rise time degradation in measured TDT pulse

Frequency Domain

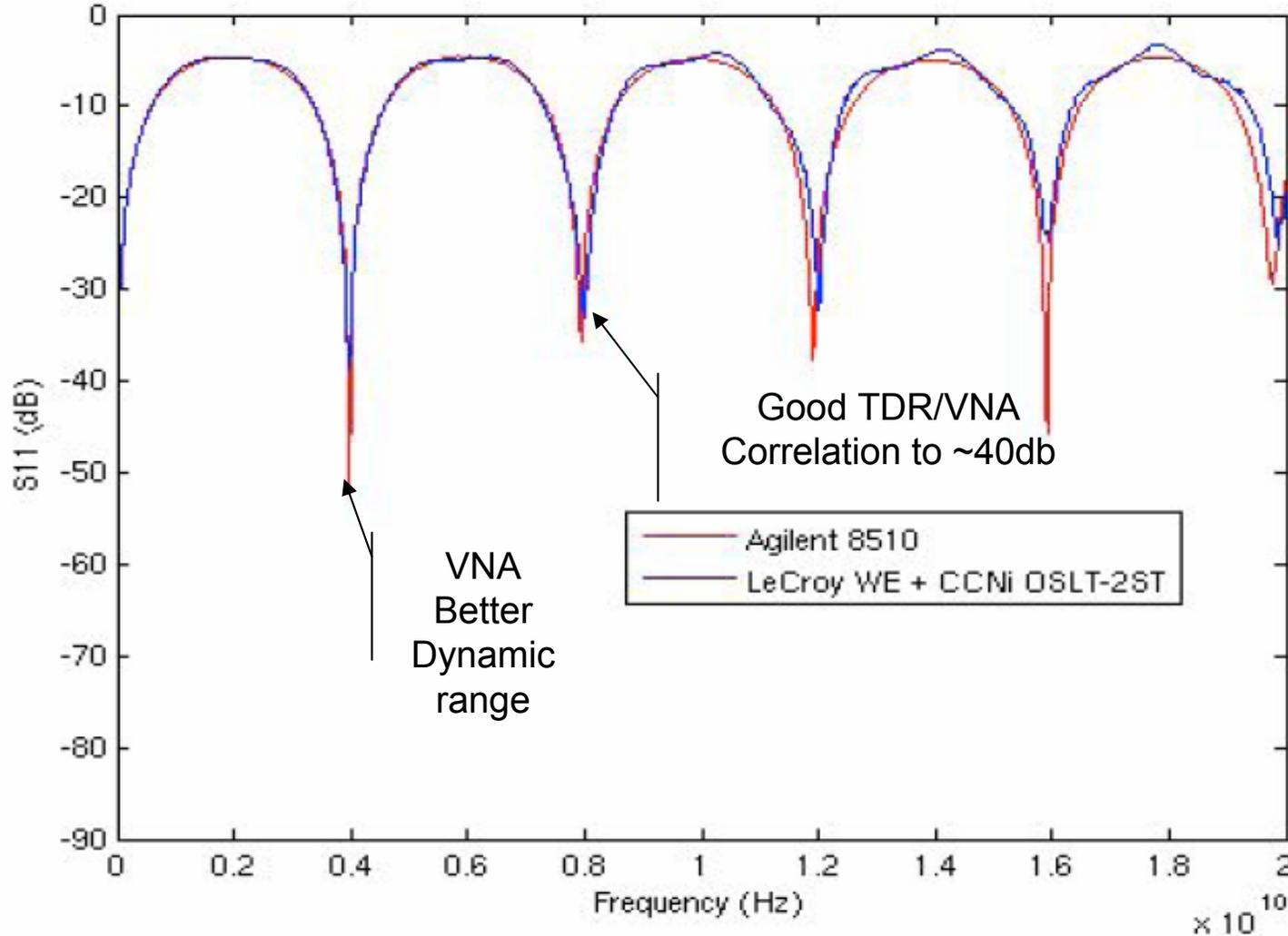
- **(S11)Return Loss** model Discontinuities
- Keep impedance ~ 10% of 50 Ohms,
- Reflected signal should be < 10% the incident signal, or the magnitude of S11 should be < 0.1. In dB, 10% is -20 dB.
- -20 dB, a common spec for S11
- **(S21)Insertion model** Dielectric, Skin Effect loss, etc.
- How much loss there is in the DUT
- If the DUT is ~ 50 Ohm impedance then > its losses, less signal will get through to the other end and the smaller S21
- - 3 dB a common spec for S21 and is a measure of the bandwidth



S-parameter Measurement Correlation

TDR/S-parameters vs. Vector Network Analyzer

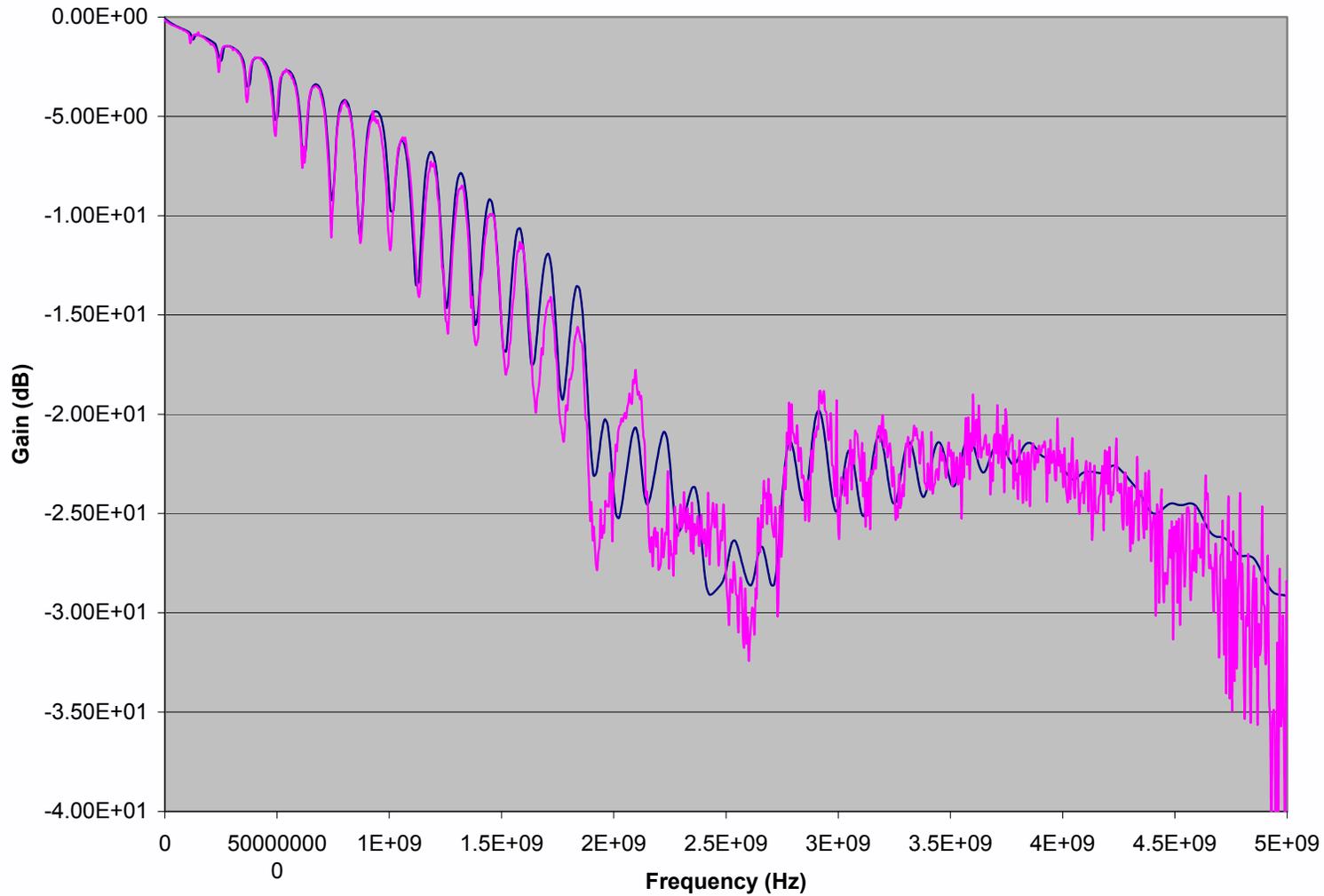
S11 Return loss Example: Measurement of terminated Beatty Standard



*Agilent 8510
VNA
compared to
WaveExpert
measurement
using Matlab*

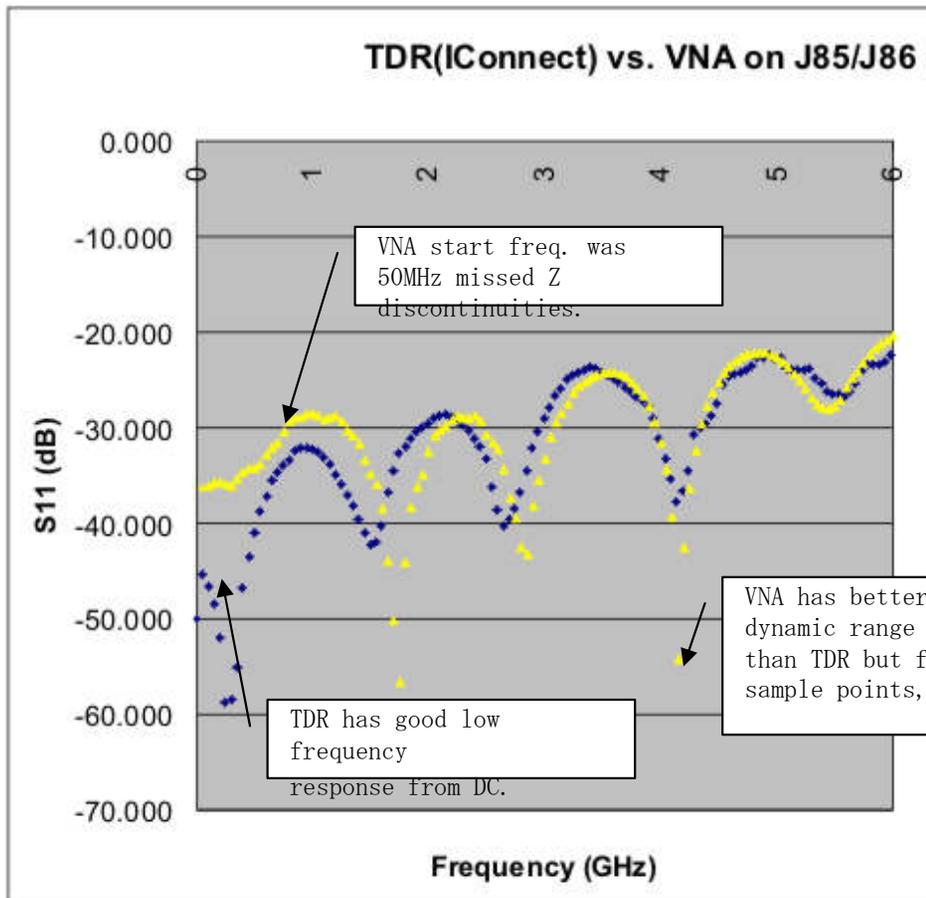
S21 of a 24-inch backplane

S21 Comparison VNA vs. TDR



TDR vs VNA

Showing DC correlation errors

6 GHz TDR to VNA Calibration Test Example

- University of New Hampshire Interoperability Laboratory (UNH) (<http://www.iol.unh.edu/>) used N1951A 20GHz PLTS VNA system to extract the s-parameters from 6 inches of etch on a SMA terminated IP test board
- S-parameters from VNA and TDR were imported into an Excel spreadsheet and plotted
- Correlation Issues:
 - TDR plotted from DC but VNA start frequency was 50MHz and did not model low frequency components
 - Another correlation issue is time domain record length. The TDR had 4000 sample points vs. Agilent of 1500 frequency points. TDR provide more Δf resolution.

Ref: DesignCon 2006 Single Port TDR Test
 For Calibrated S-Parameters by James Mayrand, & Brian Shumaker, Complete DVT Solutions

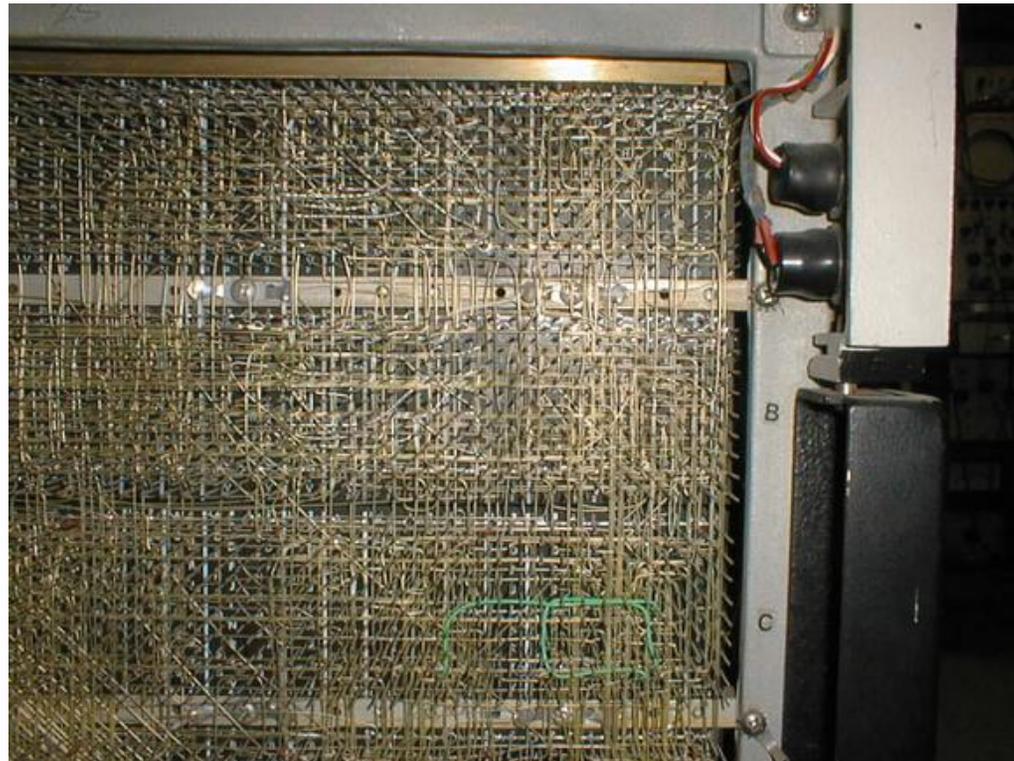
VNA vs. TDR/S-parameters

Advantages and Limitations

1. **TDR**
 1. Higher point density enables measurement of much longer devices, such as Back Planes - i.e. *better spatial resolution*
 2. Up to **100,000 pts** in any given time window Allows for *Better frequency resolution* for S-parameter measurement
 3. Ability to select a wider time window for calibration for measurement of longer devices in fixtures
 4. Able to move from time to frequency domain (TDR/T&S-Parameters)
 5. Easy to setup
 6. Fault locations (open/short) capability
 7. Less expensive

2. **VNA**
 1. Use VNA when you require more dynamic range than ~40-50dB
 2. Smaller horizontal sample points for limited time window
 3. Create complex s- parameter spice models
 4. More sensitive to calibration standards
 5. Limited TDR capability
 6. More expensive

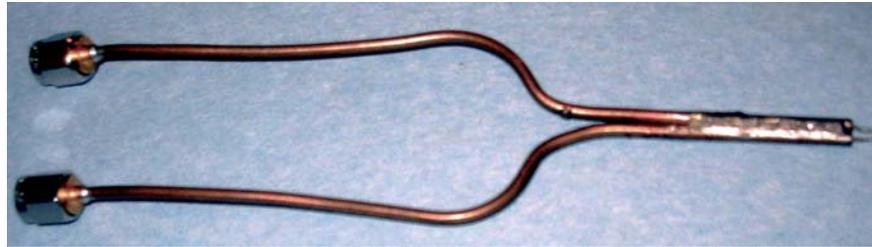
TDR Probes “When SMA's are not available”



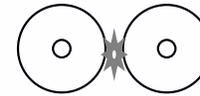
Probing Solution for TDR 50/100 ohm Hand Probe

- Low Cost
- Un-Balanced Diff
- Bandwidth/RT/FT?
- Can cause TDR Impedance and S-parameters errors
- Probe tips oxidize

Homemade Soldered 50 ohm Ridge Coax



Soldered 50ohm Cables



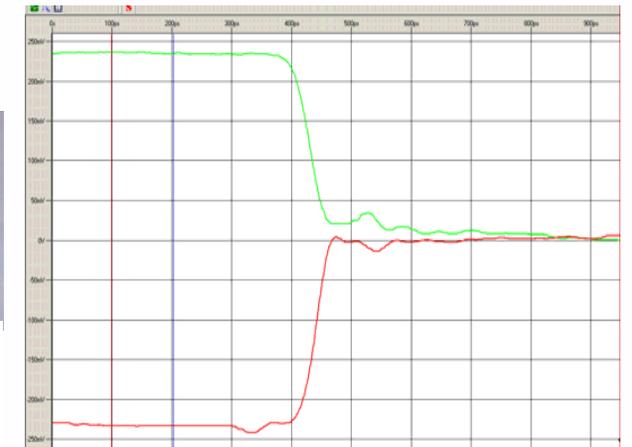
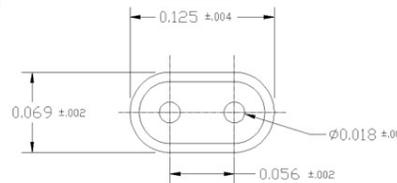
- Moderate Cost
- Balanced SMA Connector
- Balanced 100 ohm TwinAx
- Convert to 50ohm
- non oxidizing probe tips: Gold plated
- Known Rt/Ft
- More accurate Impedance and s-parameters

GigaProbes™

(www.gigaprobes.com)



Balanced 100 ohm SMA,s & TwinAx probe



Applications: Boards, Cables, Package (1mm bump side)50

Probing Solution for TDR- Gigaprobes™

DVT Solutions, LLC

Gigaprobes™



www.gigaprobes.com

Bandwidth	30 GHz
Launch Discontinuity	10mv
Pitch	.5-2mm
Probe Prop Delay	500ps
Use with Micro manipulator	Yes
Impedance	100 and 50 ohm
Accessories:	SMA wrench with .5, 1, 2mm for S-S pitch calibration
EZ-Grip Sleeves	Tie Wraps for Cable Management
Application CD	110 mm Tweezers
50 ohm Conversion kit	Manual & Datasheets
Four Right angle SMA's	Desk top Micro-Lens

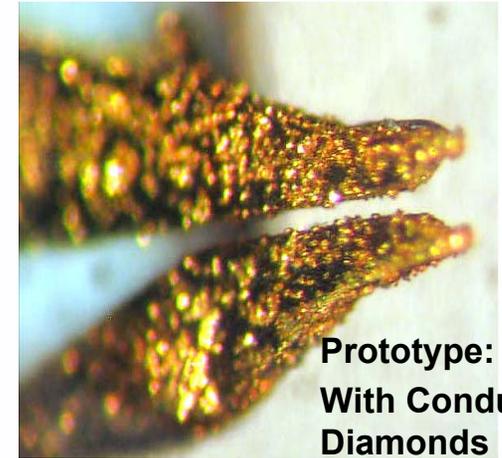
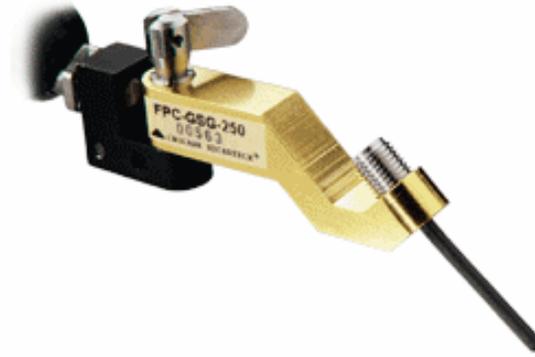
Probing Solution for TDR

MicoProbes - For Micromanipulators

- Moderate cost but high support costs: requires probe station with microscope
- Probe tips easily damaged
- High bandwidth > 40Ghz
- SLOT Calibration substrates available to de-embed probe errors
- New conductive diamond tips may improve probe tip reliability
- 50 and 100 ohm (custom) available from GGB. Ind.

Applications: Die, board, package

Cascade Microtech 50 ohm Probe



GGB Ind. Pico Probes 100 ohm Probe

